

**MICROCHIP****AN513**

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## Analog to Digital Conversion

### INTRODUCTION

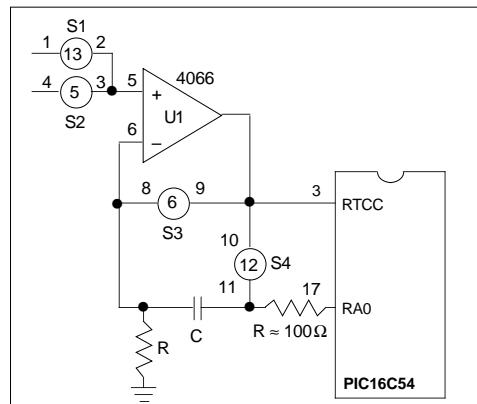
This application note describes a method for implementing analog to digital conversion on the PIC16C5X series of microcontrollers. The converter requires only five external components and is software and hardware configurable for conversion resolutions from 6 bits up to 10 bits and conversion times of 250 $\mu$ s or longer. The method is useable for both voltage and current conversion and uses a software calibration technique that compensates for time and temperature drift as well as component errors. The PIC16C5X microcontrollers are ideal for simple analog applications because:

- \* Very low cost.
- \* Few external components required.
- \* Fully programmable. PIC16C5X microcontrollers are offered as One-Time-Programmable (OTP) EPROM devices.
- \* Available off the shelf from distributors.
- \* Calibration in software for improved measurement accuracy.
- \* Power savings using PIC16C5X's Sleep mode.
- \* PIC16C5X's output pins have large, current source/sink capability to drive LED's directly.

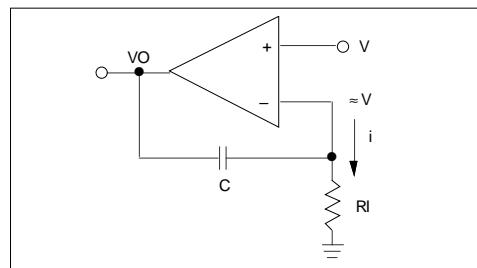
### THEORY OF OPERATION

The application uses a capacitive charging circuit (see Figure 1) to convert the input voltage to time, which can be easily measured using a microcontroller. First, the reference voltage is applied to the input voltage to current converter (U1). The equivalent circuit is shown in Figure 2. This circuit provides a linearly variable current as a function of input voltage. The logarithmic characteristic that would occur if the input voltage was applied directly to an RC is not present. The capacitor C is charged up until the threshold on the chip input trips. This generates a software calibration value that is used to calibrate out most circuit errors, including inaccuracies in the resistor and capacitor, changes in the input threshold voltage and temperature variations. After the software calibration value is measured, the capacitor is discharged (see Figure 3) and the input voltage is connected to VIN. The time to trip the threshold is measured for the input voltage and compared to the calibration value to determine the actual input voltage.

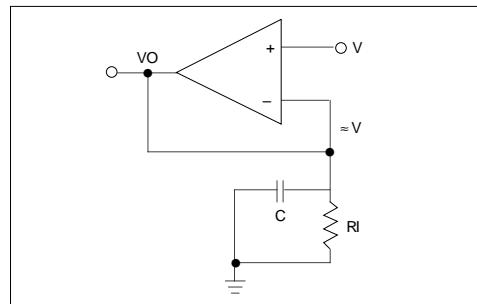
**FIGURE 1 - VOLTMETER A TO D CONVERTER**



**FIGURE 2 - VOLTMETER MEASUREMENT CYCLE**



**FIGURE 3 - VOLTMETER DISCHARGE CYCLE**



# Analog to Digital Conversion

## CIRCUIT CONFIGURATION

The values of R and C are selected based upon the number of bits of resolution required.

$$RC = (Vi \cdot T)/Vt$$

Where:

$Vi$  = Lowest voltage to be measured (at least ten lsbs')

T = Time to do the number of bits of resolution desired

$Vt$  = Threshold voltage of the PIC16C5X input being used

Actual value for RC should be slightly smaller than calculated to ensure that the PIC16C5X does not overcount during the measurement.

For example use a 3 volt input and 8 bits resolution with a 8 MHz clock and 6 instruction cycles per count:

$$Vi = 100 \text{ mV}$$

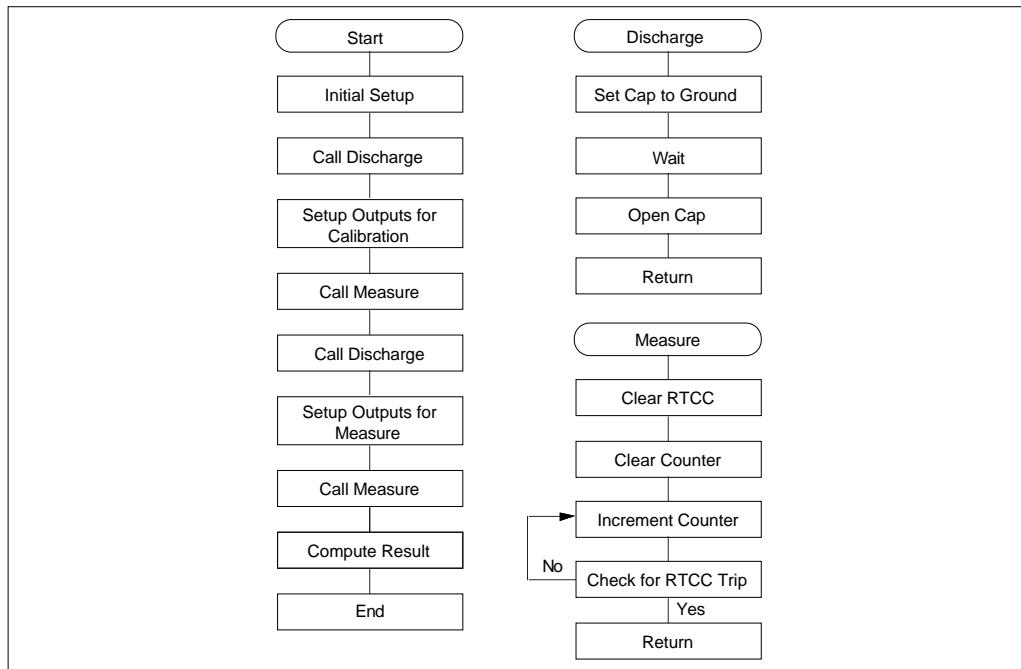
$$T = 256 * 1/8 \text{ MHz} * 4 \text{ clocks/cycle} * 6 \text{ cycles} = 768 \mu\text{s}$$

$$Vt = 3.0 \text{ V (est)}$$

For input voltages greater than 3 volts a resistor divider network should be used to keep the maximum voltage on  $Vin$  to less than 3 Volts. For best performance the reference voltage should be between 2 and 3 volts.

The circuit can also be used as a current mode A to D converter. In this case the input voltage to current converter is not needed and the reference current and input current are both routed via analog switches directly into the capacitor.

**FIGURE 4 - TRANSMISSION FLOW CHART**



## CIRCUIT PERFORMANCE

The calibration cycle removes all first order errors (offset, gain, R and C inaccuracy, power supply voltage and temperature) except the reference voltage drift. Any change in the reference voltage, including noise, between the calibration cycle and the measurement cycle may result in measurement errors. Other error sources are analog switch leakage, resistor and capacitor nonlinearities, input threshold uncertainty and time measurement uncertainty (+/- one instruction cycle time). Measured performance shows the converter to be accurate within +/- 1% of full scale.

### Example

Assembly code implementing the circuit of Figure 1 is listed in Appendix A: This code measures the time up to 16 bits and calculates the results using 16-bit multiply and divide subroutines. In actual applications, if measurement accuracy permits, it may be advantageous to use 8 bits. The math code can be substantially reduced and the measure time is reduced by the simpler code and shorter count.

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## APPENDIX A:

MPASM B0.54  
VOLTMETER/AD CONVERTER PROGRAM REV 3-29-90

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TITLE 'VOLTMETER/AD CONVERTER PROGRAM REV 3-29-90'  
LIST P=16C54,F=inhx16,n=0

0008           ACCA    EQU    8  
000A           ACCB    EQU    0A  
000C           ACCC    EQU    0C  
000E           ACCD    EQU    0E  
0010           ACCE    EQU    10  
0012           TMEAS   EQU    12  
0014           TEMP    EQU    14

0060           VCALMS  EQU    60       ;VCAL MSB VALUE IN HEX  
00A4           VCALLS  EQU    0A4     ;VCAL LSB VALUE IN HEX

ORG 1FF  
01FF 0A58      GOTO    VOLTS   ;PROGRAM CODE  
                ORG    0       ;SUBROUTINES

0000 0209      MADD    MOVF    ACCA+1,W  
0001 01EB      ADDWF   ACCB+1   ;ADD LSB  
0002 0603      BTFSF   3,0      ;ADD IN CARRY  
0003 02AA      INCF    ACCB  
0004 0208      MOVF    ACCA,W  
0005 01EA      ADDWF   ACCB     ;ADD MSB  
0006 0800      RETLW   0  
0007 0000      NOP

0008 0915      MPY     CALL    SETUP   ;RESULTS IN B(16 MSB'S) AND C(16 LSB'S)  
0009 032E      MLOOP  RRF     ACCD    ;ROTATE D RIGHT  
000A 032F      RRF     ACCD+1  
000B 0603      SKPNC  
000C 0900      CALL    MADD  
000D 032A      RRF     ACCB  
000E 032B      RRF     ACCB+1  
000F 032C      RRF     ACCC  
0010 032D      RRF     ACCC+1  
0011 02F4      DECFSZ TEMP    ;LOOP UNTIL ALL BITS CHECKED  
0012 0A09      GOTO    MLOOP  
0013 0800      RETLW   0

0014 0000      NOP  
0015 0C10      SETUP  MOVlw  10  
0016 0034      MOVWF  TEMP  
0017 020A      MOVF    ACCB,W   ;MOVE B TO D  
0018 002E      MOVWF  ACCD  
0019 020B      MOVF    ACCB+1,W  
001A 002F      MOVWF  ACCD+1  
001B 020C      MOVF    ACCC,W  
001C 0030      MOVWF  ACCE  
001D 020D      MOVF    ACCC+1,W  
001E 0031      MOVWF  ACCE+1  
001F 006A      CLRF    ACCB  
0020 006B      CLRF    ACCB+1  
0021 0800      RETLW   0

0022 0000      NOP  
0023 0915      DIV     CALL    SETUP  
0024 0C20      MOVlw  20  
0025 0034      MOVWF  TEMP  
0026 006C      CLRF    ACCC  
0027 006D      CLRF    ACCC+1

# Analog to Digital Conversion

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```

0028 0403      DLOOP    CLRC
0029 0371          RLF     ACCE+1
002A 0370          RLF     ACCE
002B 036F          RLF     ACCD+1
002C 036E          RLF     ACCD
002D 036D          RLF     ACCC+1
002E 036C          RLF     ACCC
002F 0208          MOVF    ACCA,W
0030 008C          SUBWF   ACCC,W      ;CHECK IF A>C
0031 0743          SKPZ
0032 0A35          GOTO    NOCHK
0033 0209          MOVF    ACCA+1,W
0034 008D          SUBWF   ACCC+1,W      ;IF MSB EQUAL THEN CHECK LSB
0035 0703          NOCHK   SKPC
0036 0A3E          GOTO    NOGO
0037 0209          MOVF    ACCA+1,W      ;C-A INTO C
0038 00AD          SUBWF   ACCC+1
0039 0703          BTFSZ   3,0
003A 00EC          DECF    ACCC
003B 0208          MOVF    ACCA,W
003C 00AC          SUBWF   ACCC
003D 0503          SETC
003E 036B          NOGO   RLF    ACCB+1
003F 036A          RLF    ACCB
0040 02F4          DECFSZ TEMP
0041 0A28          GOTO   DLOOP
0042 0800          RETLW   0

0043 0C0E          DSCHRG  MOVLW  B'00001110'      ;DISCHARGE C (RA0 ON)
0044 0005
0045 0CF0
0046 0034
0047 02F4          LOOP    DECFSZ TEMP
0048 0A47          GOTO   LOOP
0049 0C0F          MOVLW  B'00001111'      ;ALL RA HIGH Z
004A 0005          TRIS   5
004B 0800          RETLW   0

004C 0061          M_TIME  CLRF   1      ;CLEAR RTCC REGISTER
004D 0069          CLRF   ACCA+1
004E 0068          CLRF   ACCA
004F 03E9          TLOOP   INCFSZ ACCA+1
0050 0A54          GOTO   ENDCHK
0051 03E8          INCFSZ ACCA
0052 0A54          GOTO   ENDCHK
0053 0A56          GOTO   END_M
0054 0701          ENDCHK BTFSZ 1,0      ;CHECK FOR RTCC TRIP
0055 0A4F          GOTO   TLOOP
0056 0201          END_M  MOVF   1,W
0057 0800          RETLW   0

0058 0C06          VOLTS   MOVLW  B'00000110'      ;SET S2 AND S3 HIGH(ON WHEN ACTIVATED)
0059 0026
005A 0CF0          MOVLW  B'11110000'      ;ACTIVATE SWITCHES S1-S4
005B 0006          TRIS   6
005C 0C28          MOVLW  B'00101000'      ;SELECT POSITIVE EDGE FOR RTCC
005D 0002          OPTION
005E 0C00          MOVLW  B'00000000'
005F 0025          MOVWF  5      ;SET RA0 LOW (ON WHEN ACTIVATED)

0060 0943          MEAS    CALL   DSCHRG
0061 0C0A          MOVLW  B'00001010'      ;CHARGE CAPACITOR TO VIN
0062 0026
0063 094C          MOVWF  6      ;S2 AND S4 ON
0064 0209          CALL   M_TIME      ;MEASURE TIME
0065 0033          MOVWF  ACCA+1,W
0066 0208          MOVF   TMEAS+1      ;STORE LSB
0067 0032          MOVWF  ACCA,W
0068 0000          TMEAS   ;STORE MSB

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```
0068 0C05      CAL    MOVLW  B'00000101' ;S1 AND S3 ON
0069 0026      MOVWF  6
006A 0943      CALL   DSCHRG   ;CHARGE CAPACITOR TO VREF
006B 0C09      MOVLW  B'00001001' ;S1 AND S4 ON
006C 0026      MOVWF  6
006D 094C      CALL   M_TIME   ;MEASURE TIME

006E 0CA4      MOVLW  VCALLS
006F 002B      MOVWF  ACCB+1
0070 0C60      MOVLW  VCALMS
0071 002A      MOVWF  ACCB

0072 0908      CALL   MPY        ;MULTIPLY ACCA(TCAL) * ACCB(VREF)
0073 0213      MOVF   TMEAS+1,W
0074 0029      MOVWF  ACCA+1
0075 0212      MOVF   TMEAS ,W
0076 0028      MOVWF  ACCA

0077 0923      CALL   DIV        ;DIVIDE ACCB(TCAL * V) BY ACCA(TMEAS)

0078 0A58      GOTO VOLTS

END
```

Errors : 0  
Warnings : 0

# Analog to Digital Conversion

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**NOTES:**

