

## Resistance and Capacitance Meter Using a PIC16C622

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### INTRODUCTION

The PIC16C62X devices create a new branch in Microchip's PIC16CXX 8-bit microcontroller family by incorporating two analog comparators and a variable voltage reference on-chip. The comparators feature programmable input multiplexing from device inputs and an internal voltage reference. The internal voltage reference has two ranges, each capable of 16 distinct voltage levels. Typical applications such as appliance controllers or low-power remote sensors can now be implemented using fewer external components thus reducing cost and power consumption. The 18-pin SOIC or 20-pin SSOP packages are ideal for designs having size constraints.

The PIC16C62X family includes some familiar PIC16CXX features such as:

- 8-bit timer/counter with 8-bit prescaler
- PORTB interrupt on change
- 13 I/O pins
- Program and Data Memory

Device	Program Memory	Data Memory
PIC16C620	512 x 14	80 x 8
PIC16C621	1K x 14	80 x 8
PIC16C622	2K x 14	128 x 8

This family of devices also introduce on-chip brown-out detect circuitry and a filter on the reset input (MCLR) to the PIC16CXX mid-range microcontrollers. Brown-out Detect holds the device in reset while VDD is below the Brown-out Detect voltage of 4.0V, ± 0.2V. The reset filter is used to filter out glitches on the MCLR pin.

This application note will describe:

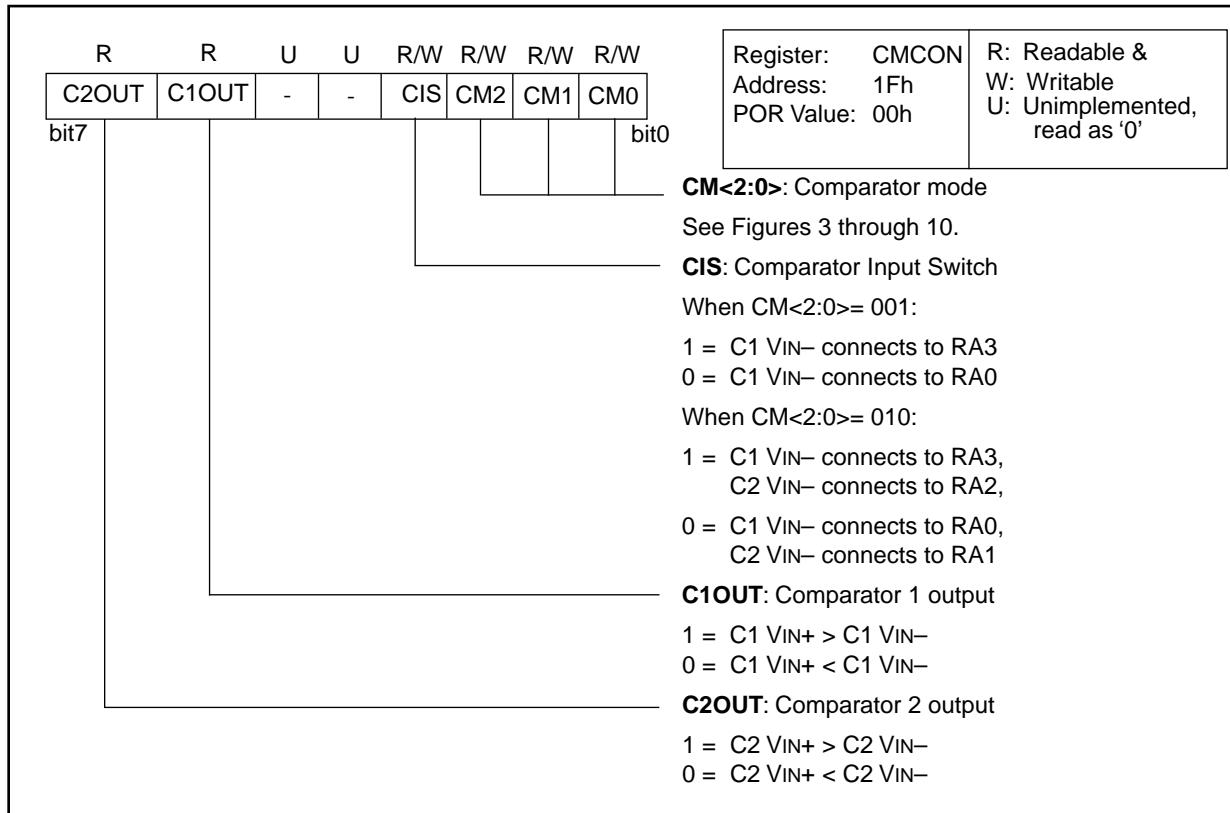
- Comparator module
  - operation
  - initialization
  - outputs
- Voltage Reference module
  - operation
  - initialization
  - outputs
- Linear slope integrating Analog to Digital conversion techniques
  - advantages
  - disadvantages
- Overview of the application circuit
- Detailed description of the measurement techniques used in the application circuit

## COMPARATOR MODULE

The comparator module contains two analog comparators with eight modes of operation. The inputs to the comparators are multiplexed with the RA0 through RA3 pins. The on-chip voltage reference can

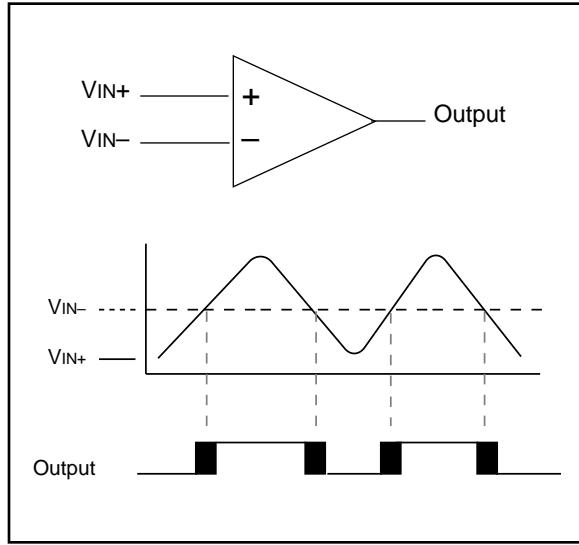
also be selected as an input to the comparators. The Comparator Control Register (CMCON) controls the operation of the comparator and contains the comparator output bits. Figure 1 shows the CMCON register.

FIGURE 1: CMCON REGISTER



A single comparator is shown in Figure 2. The relationship between the inputs and the output is also shown. When the voltage at VIN+ is less than the voltage at VIN-, the output of the comparator is at a digital low level. When the voltage at VIN+ is greater than the voltage at VIN-, the output of the comparator is at a digital high level. The shaded areas of the comparator output waveform represent the uncertainty due to input offsets and response time.

**FIGURE 2: SINGLE COMPARATOR**

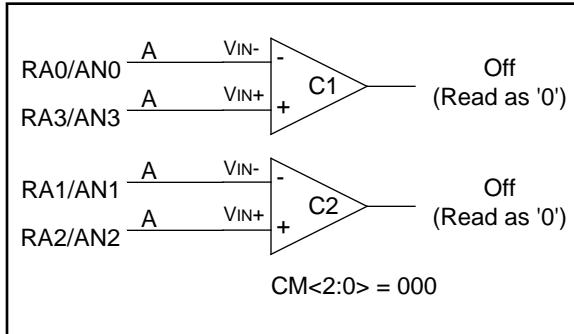


The TRISA register controls the I/O direction of the PORTA pins regardless of the comparator mode. If the comparator mode configures a pin as an analog input and the TRISA register configures that pin as an output, the contents of the PORTA data latch are placed on the pin. The value at the pin, which can be a digital high or low voltage, then becomes the input signal to the comparators. This technique is useful to check the functionality of the application circuit and the comparator module.

### Comparator Operating Modes

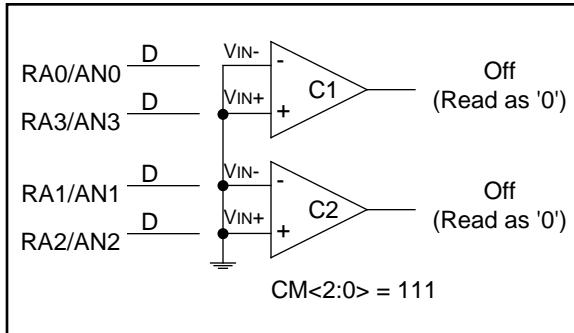
The analog inputs to the comparator module must be between Vss and Vdd and one input must be in the Common Mode Range (CMR). The CMR is defined as Vdd-1.5 volt to Vss. The output of a comparator will default to a high level if both inputs are outside of the CMR. If the input voltage deviates above Vdd or below Vss by more than 0.6 volt, the microcontroller may draw excessive current. A maximum source impedance to the comparators of 10 kΩ is recommended. Figure 3 through Figure 10 show the eight modes of operation.

**FIGURE 3: COMPARATORS RESET**



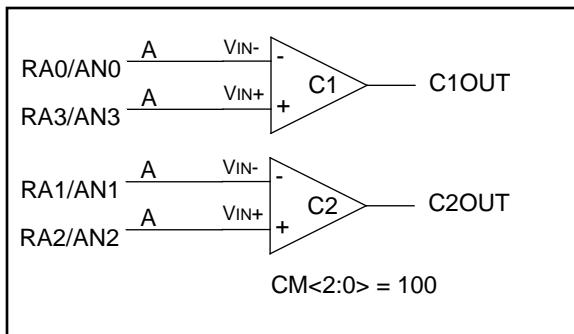
The Comparators Reset Mode (Figure 3) is considered the lowest power mode because the comparators are turned off and RA0 through RA3 are analog inputs. The comparator module defaults to this mode on Power-on Reset.

**FIGURE 4: COMPARATORS OFF**



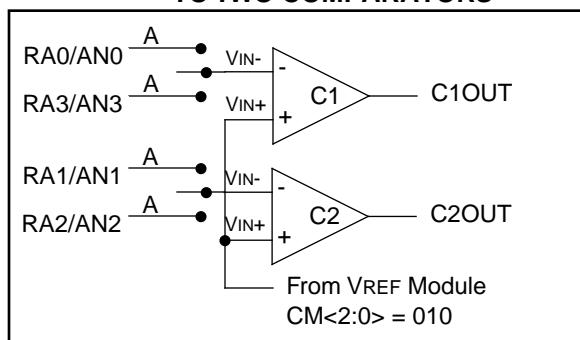
The Comparators Off Mode (Figure 4) is the same as the Comparators Reset Mode except that RA0 through RA3 are digital I/O. This mode may consume more current if RA0 through RA3 are configured as inputs and the pins are left floating.

**FIGURE 5: TWO INDEPENDENT COMPARATORS**



The Two Independent Comparators Mode (Figure 5) enables both comparators to operate independently.

**FIGURE 6: FOUR INPUTS MULTIPLEXED TO TWO COMPARATORS**

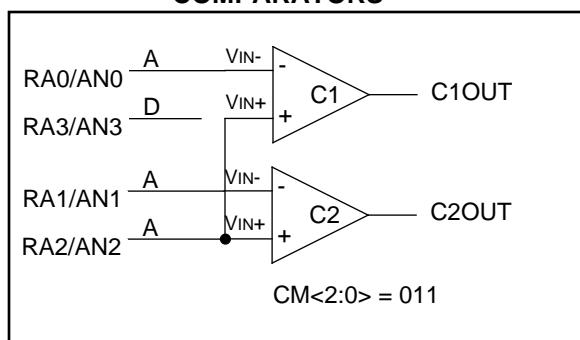


The Four Inputs Multiplexed to Two Comparators Mode (Figure 6) allows two inputs into the VIN- pin of each comparator. The internal voltage reference is connected to the VIN+ pin input of each comparator. The CIS bit, CMCON<3>, controls the input multiplexing to the VIN- pin of each comparator. Table 1 shows this relationship.

**TABLE 1: COMPARATOR INPUT MULTIPLEXING**

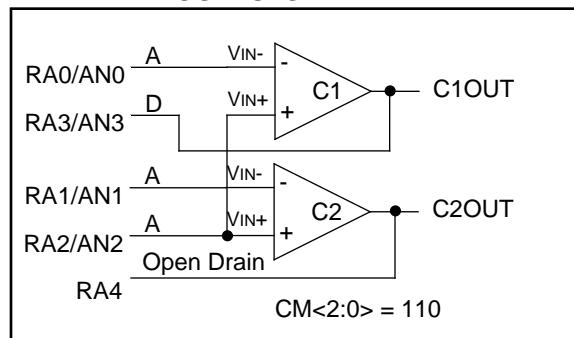
CIS	C1 VIN-	C2 VIN-
0	RA0	RA1
1	RA3	RA2

**FIGURE 7: TWO COMMON REFERENCE COMPARATORS**



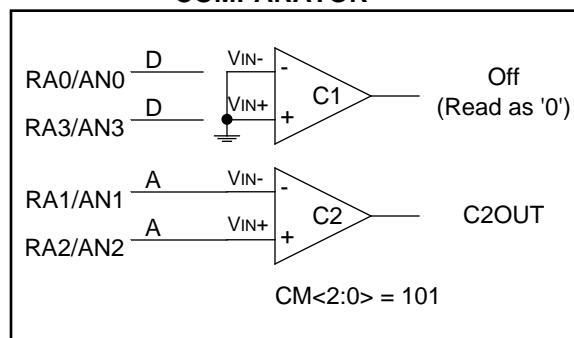
The Two Common Reference Comparators Mode (Figure 7) configures the comparators such that the signal present on RA2 is connected to the VIN+ pin of each comparator. RA3 is configured as a digital I/O pin.

**FIGURE 8: TWO COMMON REFERENCE COMPARATORS WITH OUTPUTS**



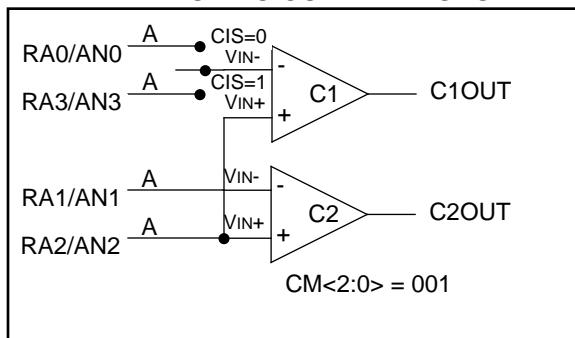
The Two Common Reference Comparators with Outputs Mode (Figure 8) connects the outputs of the comparators to an I/O pin. These outputs are digital outputs only with RA3 defined as a CMOS output and RA4 defined as an open drain output. RA4 requires a pull-up resistor to function properly. The value of resistance used for the pull-up will affect the response time of comparator C2. The signal present on RA2 is connected to the VIN+ pin of both comparators.

**FIGURE 9: ONE INDEPENDENT COMPARATOR**



The One Independent Comparator Mode (Figure 9) turns comparator C1 off making both RA0 and RA3 digital I/O. Comparator C2 is operational with analog inputs from RA1 and RA2.

**FIGURE 10: THREE INPUTS MULTIPLEXED TO TWO COMPARATORS**



The Three Inputs Multiplexed to Two Comparators Mode (Figure 10) connects the VIN+ pin of each comparator to RA2. The VIN- pin of comparator 2 is connected to RA1. The CIS bit, CMCON<3>, controls the input to the VIN- pin of comparator 1. If CIS = 0, then RA0 is connected to the VIN- pin. Otherwise RA3 is connected to the VIN- pin of comparator 1.

**Note:** Each comparator that is active will consume less power when the output is at a high level.

### Clearing the Comparator Interrupt Flag

The comparator interrupt flag, CMIF, is located in the PIR1 register. This flag must be cleared after changing comparator modes. Whenever the comparator mode or the CIS bit is changed, the CMIF may be set due to the internal circuitry switching between modes. Therefore, comparator interrupts should be disabled before changing modes. Then, a delay of 10  $\mu$ s should be used after changing modes to allow the comparator circuitry to stabilize.

The steps to clear the CMIF flag when changing modes are as follows:

- Change the comparator mode or CIS bit
- 10  $\mu$ s delay
- Read the CMCON register to end the “mismatch” condition
- Clear the CMIF bit of the PIR1 register

The value of C1OUT and C2OUT are internally latched on every read of the CMCON register. The current values of C1OUT and C2OUT are compared with the latched values, and when these values are different a “mismatch” condition occurs. The CMIF interrupt flag will not be cleared if the CMCON register has not been read.

### Using the Comparator Module

The CMCON register contains the comparator output bits C1OUT and C2OUT, CMCON<7:6>. These bits are read only. C1OUT and C2OUT follow the output of the comparators and are not synchronized to any internal clock edges. Therefore, the firmware will need to maintain the status of these output bits to determine the actual change that has occurred. The PIR1 register contains the comparator interrupt flag CMIF, PIR1<6>. The CMIF bit is set whenever there is a change in the output value of either comparator relative to the last time the CMCON register was read.

**Note:** If a change in C1OUT or C2OUT should occur when a read operation on the CMCON register is being executed (start of the Q2 pycycle), the CMIF interrupt flag may not be set.

When reading the PORTA register, all pins configured as analog inputs will read as a ‘0’. Analog levels on any pin that is defined as a digital input may cause the input buffer to consume more current than is specified.

The code in Example 1 shows the steps required to configure the comparator module. RA3 and RA4 are configured as digital outputs. RA0 and RA1 are configured as the VIN- inputs to the comparators and RA2 is the VIN+ input to both comparators.

### EXAMPLE 1: INITIALIZING THE COMPARATOR MODULE

```

CLRF    PORTA          ;init PORTA
MOVLW  0X03           ;Two Common
MOVWF  CMCON          ;Reference
                  ;Comparators
                  ;mode selected
BSF     STATUS,RP0      ;go to Bank 1
MOVLW  0X07           ;Set RA<2:0> as
MOVWF  TRISA          ;inputs,RA<4:3>
                  ;as outputs
BCF     STATUS,RP0      ;go to Bank 0
CALL   DELAY10         ;10 $\mu$ s delay
MOVF   CMCON,F         ;read the CMCON
BCF     PIR1,CMIF       ;clear the CMIF
BSF     STATUS,RP0      ;go to Bank 1
BSF     PIE1,CMIE       ;enable compar-
                  ;ator interrupt
BCF     STATUS,RP0      ;go to Bank 0
BSF     INTCON,PEIE     ;enable global
BSF     INTCON,GIE      ;and peripheral
                  ;interrupts

```

The comparators will remain active if the device is placed in sleep mode, except for the Comparators Off Mode (CM<2:0>=111) and Comparators Reset Mode (CM<2:0>=000). In these modes the comparators are turned off and are in a low power state. A comparator interrupt, if enabled, will wake-up the device from sleep in all modes except Off and Reset.

## Comparator Timings

The comparator module has a response time and a mode change to output valid timing associated with it. The response time is defined as the time from when an input to the comparator changes until the output of that comparator becomes valid. The response time is faster when the output of the comparator transitions from a high level to a low level. The mode change to output valid time refers to the amount of time it takes for the output of the comparators to become valid after the mode has changed. The internal voltage reference may contribute some delay if used in conjunction with the comparators (see Voltage Reference Settling Time).

## VOLTAGE REFERENCE MODULE

The voltage reference is a 16-tap resistor ladder network that is segmented to provide two ranges of VREF values. Each range has 16 distinct voltage levels. The voltage reference has a power-down function to conserve power when the reference is not being used. The voltage reference also has the capability to be connected to RA2 as an output. Figure 11 shows the Voltage Reference Control Register (VRCON) register which controls the voltage reference. Figure 12 shows the block diagram for the voltage reference module.

**FIGURE 11: VRCON REGISTER**

R/W	R/W	R/W	U	R/W	R/W	R/W	R/W
VREN	VROE	VRR	-	VR3	VR2	VR1	VR0
bit7							bit0

Register: VRCON  
Address: 9Fh  
POR Value: 00h

R: Readable  
W: Writable  
U: Unimplemented,  
read as '0'

**VR<3:0>:** VREF value selection  $0 \leq VR [3:0] \leq 15$

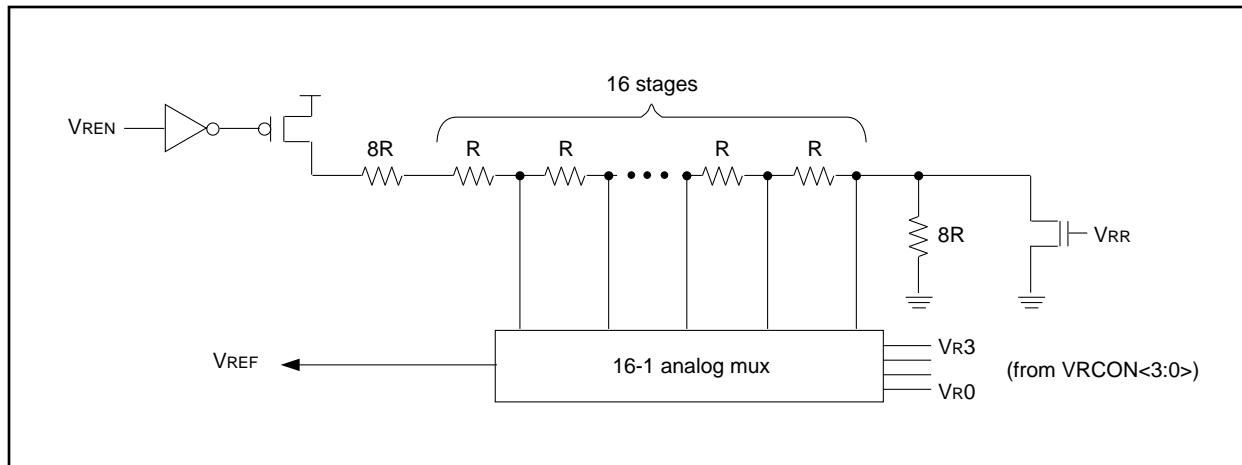
when VRR = 1:  $VREF = (VR<3:0>/ 24) * VDD$   
when VRR = 0:  $VREF = 1/4 * VDD + (VR<3:0>/ 32 * VDD)$

**VRR:** VREF Range selection  
1 = Low Range  
0 = High Range

**VROE:** VREF Output Enable  
1 = VREF is output on RA2 pin  
0 = VREF is disconnected from RA2 pin

**VREN:** VREF Enable  
1 = VREF circuit powered on  
0 = VREF circuit powered down, no IDD drain

**FIGURE 12: VOLTAGE REFERENCE BLOCK DIAGRAM**



**Note:** The voltage reference is VDD derived and therefore, the VREF output changes with fluctuations in VDD.

## Using the Voltage Reference

The voltage reference module operates independently of the comparator module. The output of the voltage reference may be connected to the RA2 pin at any time by setting the TRISA<2> bit and the VRCON<6> bit (VROE). It should be noted that enabling the voltage reference with an input signal present will increase current consumption. Configuring the RA2 pin as a digital output with the VREF output enabled will also increase current consumption. The increases in current are caused by the voltage reference output conflicting with an input signal or the digital output. The amount of increased current consumption is dependent on the setting of VREF and the value of the input signal or the digital output.

The full range of Vss to VDD cannot be realized due to the construction of the module (Figure 12). The transistors on the top and bottom of the resistor ladder network keep VREF from approaching Vss or VDD. Equation 1 and Equation 2 are used to calculate the output of the voltage reference.

### EQUATION 1: VOLTAGE REFERENCE EQUATION, VRR=1

$$V_{REF} = (VR<3:0>/24) \times V_{DD}$$

### EQUATION 2: VOLTAGE REFERENCE EQUATION, VRR=0

$$V_{REF} = (V_{DD}/4) + (VR<3:0>/32) \times V_{DD}$$

An example of how to configure the voltage reference is given in Equation 2. The reference is set for an output voltage of 1.25V at a VDD of 5.0V.

### EXAMPLE 2: VOLTAGE REFERENCE CONFIGURATION

```

MOVlw 0x02      ;4 Inputs Muxed
MOVwf CMCON     ;to 2 comps.
BSF STATUS,RP0   ;go to Bank 1
MOVLw 0x07      ;RA3-RA0 are
MOVwf TRISA     ;outputs
MOVLw 0XA6      ;enable VREF,
MOVwf VRCON     ;low range
                     ;set VR<3:0>=6
BCF STATUS,RP0   ;go to Bank 0
CALL DELAY10    ;10µs delay

```

If the voltage reference is used with the comparator module, the following steps should be followed when making changes to the voltage reference.

1. Disable the comparator interrupts
2. Make changes to the voltage reference
3. Delay 10 µs to allow VREF to stabilize
4. Delay 10 µs to allow comparators to settle
5. Clear the comparator interrupt flag
  - Read the CMCON register
  - Clear the CMIF bit
6. Enable comparator interrupts

The output of the voltage reference may be used as a simple DAC. However, the VREF output has limited drive capability when connected to the RA2 pin. In fact the amount of drive the voltage reference can provide is dependent on the setting of the tap on the resistor ladder. If VREF is used as an output, an external buffer must be utilized.

### Voltage Reference Settling Time

Settling time of the voltage reference is defined as the time it takes the output voltage to settle within 1/4 LSB after making a change to the reference. The changes include adjusting the tap position on the resistor ladder, enabling the output, and enabling the reference itself. If the voltage reference is used with the comparator module, the settling time must be considered.

## MAKING SIMPLE A/D CONVERSIONS

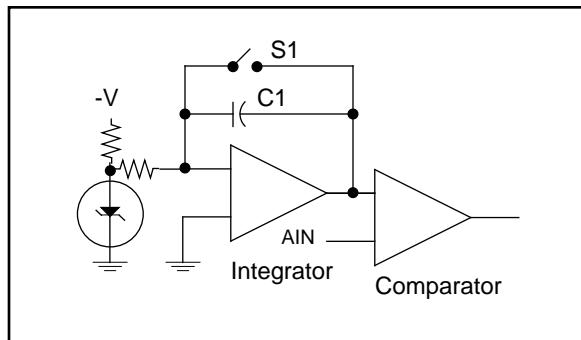
Linear slope integrating A/D converters are very simple to implement and can achieve high linearity and resolution for low conversion rates. The three types of converters that will be discussed are the single-slope, dual-slope, and modified single-slope converters. The following material was referenced from application note AN260, "A 20-Bit (1ppm) Linear Slope-Integrating A/D Converter", found in the Linear Applications Handbook from National Semiconductor®.

### Single-Slope Integrating Converter

A single-slope integrating converter is shown in Figure 13. In a single-slope converter, a linear ramp is compared against an unknown input AIN. When the switch S1 is opened the ramp begins. The time interval between the opening of the switch and the comparator changing state is proportional to the value of AIN.

The basic assumptions are that the integrating capacitor C1 and the clock used to measure the time interval remain constant over time and temperature. This type of converter is heavily dependent on the stability of the integrating capacitor.

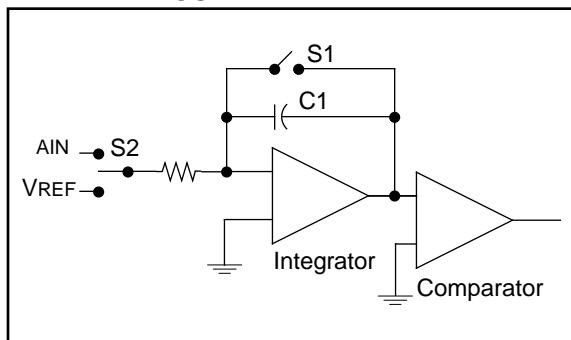
**FIGURE 13: SINGLE-SLOPE INTEGRATING CONVERTER**



### Dual-Slope Integrating Converter

Figure 14 shows a dual-slope integrating converter. The dual-slope converter integrates the AIN input for a predetermined length of time. The voltage reference is then switched into the integrator input, using S2, which integrates in a negative direction from the AIN slope. The length of time the reference slope requires to return to zero is proportional to the value of AIN. Both slopes are made with the same integrating capacitor C1 and measured with the same clock, so they need only to be stable over one conversion cycle.

**FIGURE 14: DUAL-SLOPE INTEGRATING CONVERTER**



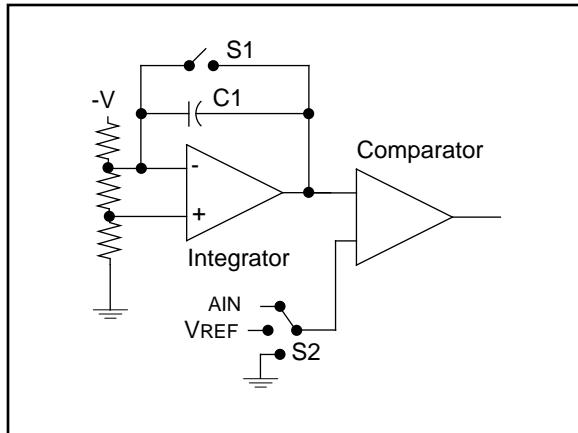
The dual-slope converter essentially removes the stability factor of the integrating capacitor from a conversion, however, the dielectric absorption of C1 has a direct effect. Dielectric absorption not only creates residual non-linearity in the dual-slope converter, but causes the converter to output different values for a fixed input as the conversion rate is varied. Dielectric absorption is defined as the capacitor dielectric's unwillingness to accept or give up charge instantaneously. This effect is modeled as a parasitic RC network across the main capacitor. A charged capacitor will require some time to discharge, even through a dead short, due to the parasitic RC network and some amount of charge will be absorbed by the parasitic C after charging of the main capacitor has stopped. Typically, Teflon, polystyrene and polypropylene dielectrics offer better performance than paper, mylar, or glass. Electrolytics have the worst dielectric absorption characteristics and should be avoided for use in slope integrating converters.

## Modified Single-Slope Converter

The modified single-slope converter has been designed to compensate for the effects present in the previous converters. Resolutions of up to 16-bits can be achieved using high precision components and voltage reference source. Figure 15 shows the modified single-slope converter. Some features of this converter are:

- Continuously corrects for zero and full-scale drifts in all components of the circuit.
  - The integrating capacitor C1 is charged periodically and always in the same direction. The error induced from dielectric absorption will be small and can be compensated by using an offset term in the calibration procedure.
  - The ramp voltage always approaches the comparator trip point from the same direction and slew rate.
  - There is no noise rejection capability because the input signal is directly coupled to the comparator input. A filter at the comparator input would cause a delay due to the settling time of the filter.

**FIGURE 15: MODIFIED SINGLE-SLOPE  
INTEGRATING CONVERTER**



The microcontroller sends a periodic signal to the switch S1 regardless of the operating mode of the system. The output of the integrator is a fixed frequency, period and height signal which is fed into the input of the comparator. The time between ramps is long enough to allow the integrating capacitor C1 to discharge completely. The other input is multiplexed with ground, reference, and the AIN through switch S2. When the microcontroller starts a conversion, the ground signal is switched into the comparator and the time for the ramp to cross zero is measured and stored. The same measurements are repeated for the reference and AIN signals. Assuming that the integrator ramps are highly linear, Equation 3 is used to determine the value of AIN.

**EQUATION 3: OUTPUT EQUATION FOR THE MODIFIED-SLOPE CONVERTER**

$$AIN = \frac{\tau_{AIN} - \tau_{GND}}{\tau_{VREF} - \tau_{GND}} \times K \text{ } \mu V$$

where  $\tau_{AIN}$  is the measured time for the AIN signal,  $\tau_{VREF}$  is the measured time for the voltage reference signal,  $\tau_{GND}$  is the measured time for the ground signal, and K is a constant (typically  $10^7$ ).

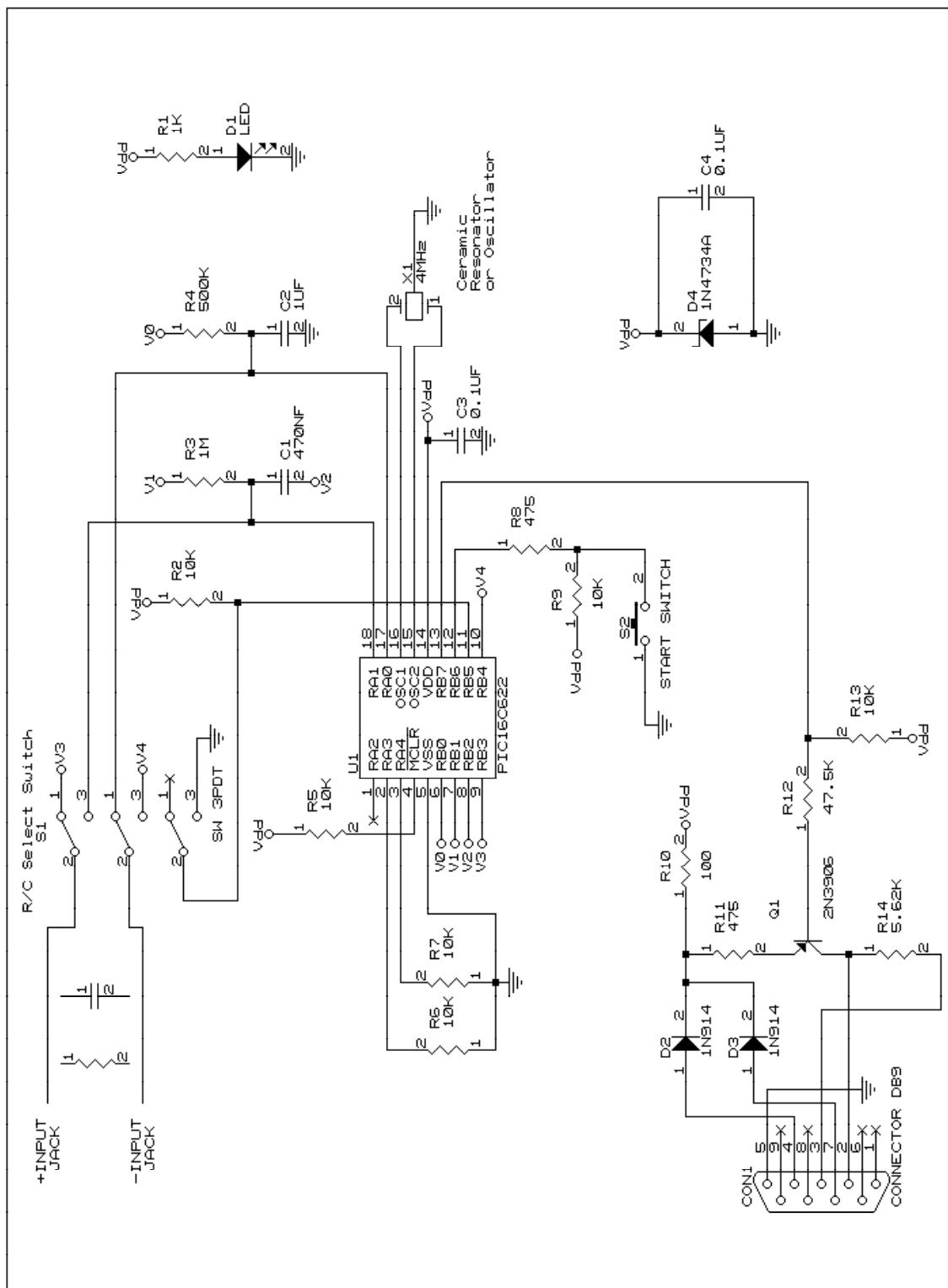
## APPLICATION CIRCUIT

The application circuit, called PICMETER, uses a PIC16C622 as a resistance and capacitance meter. The PICMETER uses a variation of the single-slope integrating convertor. The linear slope and integrator of Figure 13 are replaced with the exponential charge waveform of a RC Network. The charge time of a known component is compared against the charge time of an unknown component to determine the value of the unknown component.

A schematic of the PICMETER is shown in Figure 16. All reference designators cited in this section refer to this schematic. Results are transmitted to a PC which displays the value measured. The PICMETER can measure resistance in the range 1 K $\Omega$  to 999 K $\Omega$  and capacitance from 1 nF to 999 nF.

The following sections describe, in detail, the hardware, firmware, and PC software used in the application circuit. Appendix A shows the PICMETER firmware and Appendix B has the PC software. Appendix C contains the PCB layout.

**FIGURE 16: PICMETER SCHEMATIC**



## Power

The RS-232 serial port provides power to the PICMETER. The RTS and DTR lines from the serial port output 3V to 11V to the PICMETER. The diodes D2 and D3 prevent any damage to the PC's serial port. Resistor R10 is used to current limit the Zener diode, D4. D4 is used to regulate the RTS and DTR voltage to 5.6V. Capacitors C3 and C4 provide power supply filtering to the Zener diode and the PIC16C622. This method of supplying power to devices using a serial port, such as a trackball or mouse, is very simple considering that the PICMETER requires approximately 7 mA to function.

## Switches

Switch S1 is used to select either a resistor or capacitor measurement. RB5 of the PIC16C622 is used to detect what type of component is being measured. This switch also swaps the unknown component into the RC network.

If a resistor is the unknown component and a capacitor measurement is requested, the circuit reduces to a resistor divider on the VIN- pin of the comparator. This would result in a measured value of 0 pF if the voltage on the resistor divider network is greater than the voltage reference setting. Otherwise an error is detected. If a capacitor is the unknown component and a resistor measurement is selected, the circuit reduces to a capacitor divider network on the VIN- pin of the comparator. This case will also produce an error message.

Resistor measurements that are started without any component connected to the measuring terminals will cause an error. Capacitor measurements without a component connected to the measuring terminals will give a result of 0 pF.

Switch S2 is used to initiate a measurement. The switch is connected to RB6 of the PIC16C622 and the PORTB wake-up on change interrupt is used to detect a key press. A modified version of the firmware in AN552, "Implementing Wake-up on Key Stroke" was used to control the interrupt.

## Measuring the Charge Time

The procedures for measuring a resistor or capacitor are the same except for the I/O pins used to control the RC networks. This also applies when measuring a known or unknown component.

### Measurement Overview

The charge time of the unknown RC network is measured using Timer0. This value is multiplied by the known value of resistance or capacitance and stored in an accumulator. Then the charge time of the known RC network is measured. The accumulator is divided by the known RC network charge time to give the value of resistance or capacitance of the unknown component. Equation 4 shows the equation used to calculate resistance and Equation 5 shows the capacitance equation.

### **EQUATION 4: RESISTANCE EQUATION**

$$R_{UNK} = \frac{\tau_{UNK} \times R_{KN}}{\tau_{KN}}$$

### **EQUATION 5: CAPACITANCE EQUATION**

$$C_{UNK} = \frac{\tau_{UNK} \times C_{KN}}{\tau_{KN}}$$

RUNK and CUNK are the unknown resistor or capacitor values. RKN and CKN are the known resistor and capacitor values.  $\tau_{UNK}$  and  $\tau_{KN}$  are the charge times for the unknown and known components.

## Detailed Measurement Description

The first step in measuring the charge time of either the known or the unknown RC networks is to reconfigure the I/O pins. The default state of the PORTA and PORTB pins connected to the RC network are all grounded outputs. This discharges all capacitors in the RC networks. The unknown component is measured first, so the known component, R4 or C1, is removed from the RC network. This is accomplished by making RB0 or RB2 on the PIC16C622 an input. Connections to the other RC network are kept grounded.

The analog modules are now initialized. The mode of the comparators is set to Four Inputs Multiplexed to Two Comparators (Figure 6). The CIS bit, CMCON<3> is cleared to select RA0 as the VIN- input to comparator 1 and RA1 as the VIN- input to comparator 2. The voltage reference is enabled, the output is disabled, and the high range is selected. The tap on the resistor ladder is set to 12. The value of 12 was selected because it is the lowest value of VREF that will trip the comparators, yet gives a time constant long enough to achieve good resolution for the measurement. After a 20 msec delay, which allows the analog modules to stabilize, the comparator flag is cleared. Comparator interrupts are enabled and Timer0 is cleared. Finally, the PEIE bit is set to enable comparator interrupts and the GIE bit is set to enable interrupts.

Now that the analog systems are ready, Timer0 is cleared again and power is applied to the unknown RC network by setting RB1 or RB3 high. Timer0 begins to increment a set of three registers which are cascaded together. These registers contain the charge time of the component. While waiting for the DONE flag, the ERROR flag is checked. See the Error Message section for an explanation of error detection. When the capacitor voltage trips the comparator, Timer0 is prevented from further incrementing the time registers and the DONE flag is set. The value in the time registers is  $\tau_{UNK}$ .

The analog modules are now disabled. The comparator interrupts are disabled and the comparators are turned off ( $CM<2:0>=111$ ). RA0 through RA3 and RB0 through RB4 are set up as grounded outputs to discharge the capacitors in the RC networks. This prevents a false reading during the next measurement. The voltage reference is disabled to conserve power and all interrupt flags are cleared. Extra delay loops are added at this time to ensure that the capacitors are discharged.

The charge time,  $\tau_{UNK}$ , is then multiplied by the value of known resistance or capacitance. These values, in pF or  $\Omega$ , were obtained by measuring the known RC networks with a Fluke meter. Each of these values is a 24-bit number. The result of multiplication is a 56-bit number which is stored in accumulators ACCb (most significant 24-bits) and ACCc (least significant 24-bits).

The process now repeats itself, except this time the charge time of the known RC network is measured. Now the unknown component is removed from the RC network by making the connections from the PIC16C622 inputs. The analog modules are initialized and the same procedure explained above is followed to measure the charge time of the known RC network. The 56-bit result previously stored in accumulators ACCb and ACCc is now divided by the charge time of the known component,  $\tau_{KN}$ . This result is a 24-bit number which has the units of pF or  $\Omega$ . This value is then transmitted to the PC.

## RS-232 Transmission

PICMETER uses a transmit only, software implemented serial port adapted from AN593, "Serial Port Routines Without Using the RTCC". Hardware hand-shaking is not used. Since the serial port is realized in software, all interrupts must be disabled during transmission otherwise the baud rate can get corrupted.

On power-up, PICMETER sends a boot message to the PC which is "PICMETER Booted!". Otherwise, a four byte packet structure with a command byte and 3 data bytes is used. The command byte contains one of four possible commands:

- ASCII 'S' signifies that a measurement has been initiated
- ASCII 'E' tells the PC that an error has been detected
- ASCII 'R' tells the PC that resistance data is contained in the three data bytes
- ASCII 'C' tells the PC that capacitance data is contained in the three data bytes

The first data byte for the 'R' and 'C' commands contain the MSB of the measured value. The last data byte contains the LSB of the measured value. The three data bytes for the commands 'S' and 'E' do not contain any useful information at this time.

An 'S' command is issued every time the start switch, S2, is pressed. PICMETER then sends an 'R' or 'C' command for a valid measurement or an 'E' command when an error is detected.

Since the PICMETER operates from a single supply voltage, a discrete transistor is used as a level shifter. This insures that a low output on the RS-232 TXD line is between -3V and -11V. When the TXD line, RB7, from the PIC16C622 is at a logic high level, the transistor Q1 is off. The RXD line of the computer will then be at approximately the same voltage as the TXD line, -11V to -3V. A logic low level from RB7 of the PIC16C622 will turn on transistor Q1. This will bring the RXD line of the computer to about the same voltage of the DTR or RTS line, +3V to +11V.

The pins of interest on the DB9 connector CON1 are:

- pin 2 - RXD
- pin 3 - TXD
- pin 4 - DTR
- pin 5 - GND
- pin 7 - RTS

RTS, DTR, and GND provide power and ground to the PICMETER. RXD is connected to the collector of transistor Q1. TXD is connected to RXD through resistor R14. Since hardware hand-shaking is not implemented on the PICMETER, DSR (pin 6) and CTS (pin 8) are left disconnected.

The demo board developed by Microchip was intended to connect directly to a 9-pin serial port. A 9-pin male-to-female cable may also be used. These boards were manufactured by Southwest Circuits located in Tucson, Arizona (Appendix C). The PCB layout for this demo board is shown in Appendix C.

## Error Message

The error message is sent only when the PICMETER is making a measurement and detects an error. The range of resistance that the PICMETER measures is 1 k $\Omega$  to 999 k $\Omega$ . Using the value of C2, 1  $\mu$ F, the range of charging times for resistance measurements is 1msec to 999 ms. The range of capacitor charging times is also 1 ms to 999 ms using the resistance value of R3, 1 M $\Omega$ , and a capacitor measuring range of 1 nF to 999 nF. A ceramic resonator of 4 MHz gives Timer0 a resolution of 1  $\mu$ sec. Therefore, the highest count that the time registers should reach is 999,000. This is a 20-bit number. If the 21<sup>st</sup> bit should ever be set, it is assumed that the PICMETER is trying to measure the air gap between the measuring terminals, a component that is out of range, or switch S1 is not set correctly for the component in the measuring terminals.

## 24-Bit Math Routines

The 24-bit math routines were developed using simple algorithms found in any computer math book. These math routines include addition, subtraction, multiplication, division, and 2's complement. Four 24-bit accumulators located in the general purpose RAM area of the PIC16C622 are used by the math routines: ACCa, ACCb, ACCc, and ACCd. Table 2 shows the relationship between the math routines and the accumulators.

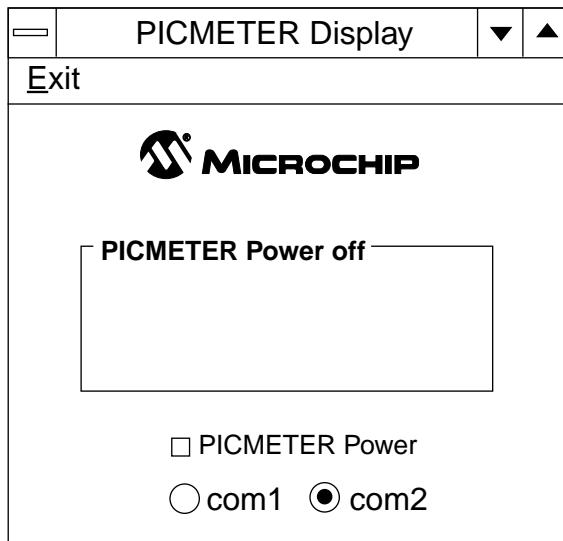
**TABLE 2: MATH ROUTINE  
ACCUMULATORS**

Name	Operation	Result	Temp. Storage
Add	ACCa + ACCb	ACCb	N/A
Subtract	2's Comp ACCa then	ACCa	N/A
	ACCa + ACCb	ACCa	
Multiply	ACCa x ACCb	ACCb (MSB's) ACCc (LSB's)	ACCd
Divide	ACCb:ACCc ACCa	quotient in ACCc remainder in ACCb	ACCd
2's Comp	NOT(ACCa) + 1	ACCa	N/A

## Computer Program

The program that receives data from the PICMETER was written in Visual Basic® from Microsoft® for the Windows® environment. Figure 17 show the display of the Windows based PICMETER program.

**FIGURE 17: PICMETER PC PROGRAM**



The operation of this program is simple. A functional description is given below:

- a) Select the appropriate COM port by clicking on the COM1 or COM2 buttons.
- b) Turn power on to the PICMETER by clicking on the PICMETER Power button.
- c) The frame message should read "PICMETER Booted!", the frame contents will be cleared, and the LED on the PICMETER should be on.
- d) The switch S1 selects the type of component that is in the measuring terminals.
- e) Pressing the START button, S2, on the PICMETER will initiate a measurement. The frame message should read "Measuring Component" and the contents of the frame will be cleared.
- f) When the measurement is complete, the frame message will read "Resistance" or "Capacitance" depending on the position of switch S1. The value of the component will be displayed in the frame as well as the units.
- g) If an error is detected, the frame message will read "Error Detected". This is only a measurement error. Check the component on the measuring terminals and the position of switch S1.
- h) Turn off the PICMETER by clicking on the PICMETER Power button. The frame message will change to "PICMETER Power OFF", the frame contents will be cleared, and the LED on the PICMETER will turn off.

Appendix B contains a complete listing of the Visual Basic program.

## PICMETER ACCURACY

The PICMETER measures capacitance in the range of 1 nF to 999 nF. Table 3 shows a comparison of various capacitors. All capacitors have a tolerance of 10% and have various dielectrics. The average error percentage is 3%.

**TABLE 3: CAPACITANCE MEASUREMENTS**

Capacitance Accuracy			
Marked Value	Fluke Value	PICMETER Value	Error %
2.2 nF	2.3 nF	2.2 nF	4.3
2.5 nF	2.63 nF	2.5 nF	4.9
20 nF	16.5 nF	16.3 nF	1.2
33 nF	35.2 nF	35.8 nF	1.7
47 nF	45 nF	44.5 nF	1.1
50 nF	52 nF	52.9 nF	1.7
100 nF	99.7 nF	93 nF	6.7
0.1 $\mu$ F	95 nF	96.1 nF	1.2
0.1 $\mu$ F	99.4 nF	102.8 nF	3.4
0.22 $\mu$ F	215 nF	215.2 nF	0.1
470 nF	508 nF	518.9 nF	2.1
940 nF	922 nF	983.1 nF	6.6

The 2.5 nF, 100 nF and 940 nF capacitors all have polyester dielectric material. The Equivalent Series Resistance (ESR) of polyester capacitors is typically high which would cause the PICMETER to have a larger error than other dielectrics. If the error percentages for these capacitors is ignored, the average error decreases to 1.9%.

The resistance range of the PICMETER is 1 k $\Omega$  to 999 k $\Omega$ . Table 4, Resistance Measurements, shows a comparison of various resistors in this range. All resistors have a tolerance of 5%. The average error percentage is 1%.

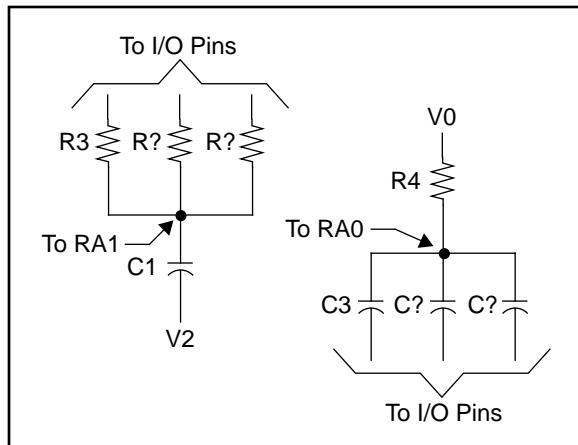
**TABLE 4: RESISTANCE MEASUREMENTS**

Resistance Accuracy			
Marked Value	Fluke Value	PICMETER Value	Error %
1.2K	1.215K	1.2K	1.3
5.1K	5.05K	5.0K	1.0
8.2K	8.47K	8.3K	2.0
10K	10.2K	10K	2.0
15K	15.36K	15.1K	1.7
20K	20.8K	20.5K	1.5
30K	30.4K	30K	1.4
51K	50.3K	49.8K	1.0
75K	75.5K	74.8K	1.0
91K	96.4K	95.9K	0.6
150K	146.3K	145.6K	0.5
200K	195.5K	195K	0.3
300K	309K	309.5K	0.2
430K	433K	434.5K	0.4
560K	596K	599.6K	0.6
680K	705K	709.8K	0.7
820K	901K	907.3K	0.7
910K	970K	977.8K	0.8

# AN611

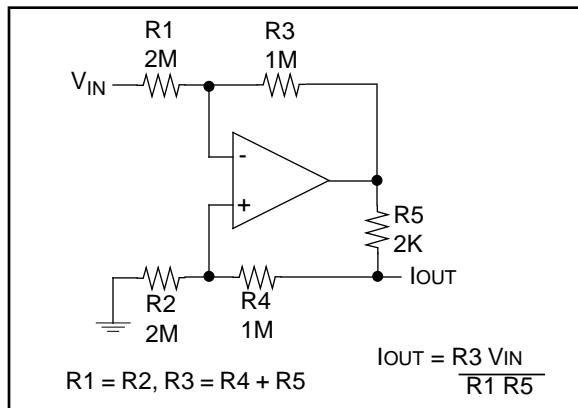
The accuracy of the PICMETER is dependent on the range of components being measured. If auto-ranging could be implemented, the accuracy of the PICMETER could be improved. The addition of capacitors in parallel with C2 of Figure 16 would allow auto-ranging for resistor measurements. Additional resistors in parallel with R3 would give auto-ranging capability to capacitor measurements. Figure 18 shows a simple implementation of auto-ranging given that the I/O pins are available. The R? and C? are the extra components that are added to the PICMETER circuit. These components should be optimized for a particular range of devices.

**FIGURE 18: AUTO-RANGING TECHNIQUE**

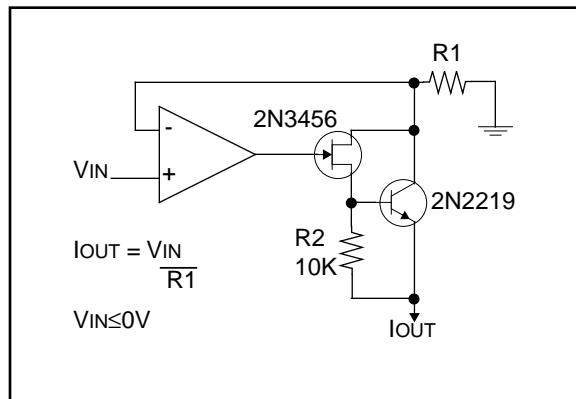


Another addition to the PICMETER that would increase the accuracy of components being measured is a constant current source. The source would feed into the resistor of the RC networks. This provides the same charging current to all RC networks being measured. Figure 19 shows a bilateral current source and Figure 20 shows a precision current source.

**FIGURE 19: BILATERAL CURRENT SOURCE**

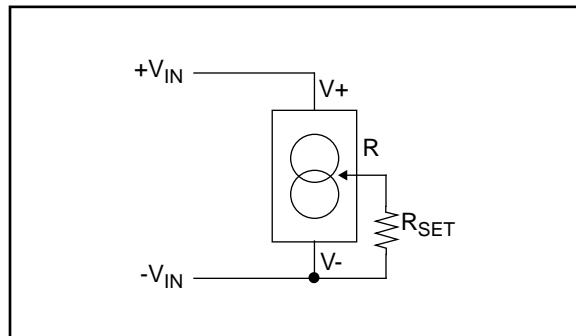


**FIGURE 20: PRECISION CURRENT SOURCE**



The alternative to the previous current sources is a single chip solution. A 3-terminal adjustable current source, such as a LM134/LM234/LM334 from National Semiconductor, is an ideal choice. This output current is programmable from 1 μA to 10 mA and requires a single external resistor to set the value of current. Figure 21 shows a block diagram of the LM334Z.

**FIGURE 21: LM334Z BLOCK DIAGRAM**



## CONCLUSION

The PIC16C62X devices add two significant analog features to the PIC16CXX mid-range family: comparators and a voltage reference. The flexibility of eight operating modes for the comparator module allows the designer to tailor the PIC16C62X device to the application. The addition of an on-chip voltage reference simplifies the design by removing at least one external component and power consumption. These analog modules coupled with the PIC16CXX mid-range family core create a new path to achieve high resolution results.

Please check the Microchip BBS for the latest version of the source code. For BBS access information, see Section 6, Microchip Bulletin Board Service, page 6-3.

## APPENDIX A: PICMETER FIRMWARE

MPASM 01.02.05 Intermediate PICMETER.ASM 5-1-1995 11:29:17  
PICMETER Firmware for PIC16C622

PAGE 1

LOC	OBJECT CODE	LINE SOURCE TEXT
	VALUE	
		0001        TITLE "PICMETER Firmware for PIC16C622"
		0002        LIST P = 16C622, F = INHX8M
		0003
		0004        INCLUDE "C:\PICMASTR\P16CXX.INC"
		0002 ; P16CXX.INC Standard Header File, Version 0.2 Microchip Technology, Inc.
		0004
		0005
3FB9		0006        FUSES _BODEN_OFF&_CP_OFF&_PWDT_ON&_WDT_OFF&_XT_OSC
		0007
		0008 ;*****
		0009 ;-----*
		0010 ;*-
		0011 ;*-      PICMETER - Resistance and Capacitance Meter
		0012 ;*-
		0013 ;*-
		0014 ;*-
		0015 ;*-      Author:      Rodger Richey
		0016 ;*-                  Applications Engineer
		0017 ;*-      Filename:    picmtr.asm
		0018 ;*-      Revision:    1 May 1995
		0019 ;*-
		0020 ;*-
		0021 ;*-
		0022 ;*-      PICMETER is based on a PIC16C622 which has two comparators and
		0023 ;*-      a variable voltage reference. Resistance and capacitance is
		0024 ;*-      calculated by measuring the time constant of a RC network. The
		0025 ;*-      toggle switch selects either resistor or capacitor input. The
		0026 ;*-      pushbutton switch starts a measurement. The time constant of the
		0027 ;*-      unknown component is compared to that of known component to
		0028 ;*-      calculate the value of the unknown component. The following
		0029 ;*-      formulas are used:
		0030 ;*-
		0031 ;*-      Resistance:    Ru = ( Rk * Tu ) / Tk
		0032 ;*-      Capacitance:   Cu = ( Ck * Tu ) / Tk
		0033 ;*-
		0034 ;*-
		0035 ;*****
		0036
		0037
		0038 ;*****
		0039 ;-----*
		0040 ;*-      RS232 code borrowed from Application Note AN593
		0041 ;*-      "Serial Port Routines Without Using the RTCC"
		0042 ;*-      Author: Stan D'Souza
		0043 ;*-
		0044 ;*****
003D 0900		0045 xtal    equ     .4000000
2580		0046 baud    equ     .9600
000F 4240		0047 fclk    equ     xtal/4
		0048 ;*****
		0049 ;The value baudconst must be a 8-bit value only
0020		0050 baudconst    equ     ((fclk/baud)/3-2)
		0051 ;*****
		0052
		0053

```

0054 ;*****
0055 ;      Bit Equates
0056 ;*****
0000 0057 BEGIN equ 0           ;begin a measurement flag
0007 0058 DONE equ 7          ;done measuring flag
0005 0059 WHICH equ 5         ;R or C measurement flag
0003 0060 F_ERROR equ 3        ;error detection flag
0005 0061 EMPTY equ 5         ;flag if component is connected
0000 0062 V0   equ 0           ;power for R reference ckt
0001 0063 V1   equ 1           ;power for C reference ckt
0002 0064 V2   equ 2           ;ground for C reference ckt
0003 0065 V3   equ 3           ;power for unknown R ckt
0004 0066 V4   equ 4           ;ground for unknown C ckt
0007 0067 msb_bit equ 7        ;define for bit 7
0000 0068 lsb_bit equ 0        ;define for bit 0
0007 0069 RkHI equ 0x07       ;value of the known resistance, R4, in ohms
009D 0070 RkMID equ 0x9D      ;measured by a Fluke meter
0038 0071 RkLO equ 0x38
0007 0072 CkHI equ 0x07       ;value of the known capacitance, C1, in pF
00C8 0073 CkMID equ 0xC8      ;measured by a Fluke meter
0030 0074 CkLO equ 0x30
0075
0076 ;*****
0077 ;      User Registers
0078 ;*****
0079 ;      Bank 0
0020 0080 W_TEMP equ 0x20      ;Bank 0 temporary storage for W reg
0021 0081 STATUS_TEMP equ 0x21  ;temporary storage for STATUS reg
0023 0082 Ttemp equ 0x23       ;temporary Time register
0024 0083 flags equ 0x24       ;flags register
0025 0084 count equ 0x25       ;RS232 register
0026 0085 txreg equ 0x26       ;RS232 data register
0027 0086 delay equ 0x27       ;RS232 delay register
0028 0087 offset equ 0x28      ;table position register
0029 0088 msb   equ 0x29       ;general delay register
002A 0089 lsb   equ 0x2A       ;general delay register
0040 0090 TimeLO equ 0x40      ;Time registers
0041 0091 TimeMID equ 0x41
0042 0092 TimeHI equ 0x42
0093
0094 ;      Math related registers
0050 0095 ACCaHI equ 0x50       ;24-Bit accumulator a
0051 0096 ACCaMID equ 0x51
0052 0097 ACCaLO equ 0x52
0053 0098 ACCbHI equ 0x53       ;24-Bit accumulator b
0054 0099 ACCbMID equ 0x54
0055 0100 ACCbLO equ 0x55
0056 0101 ACCcHI equ 0x56       ;24-Bit accumulator c
0057 0102 ACCcMID equ 0x57
0058 0103 ACCcLO equ 0x58
0059 0104 ACCdHI equ 0x59       ;24-Bit accumulator d
005A 0105 ACCdMID equ 0x5A
005B 0106 ACCdLO equ 0x5B
005C 0107 temp   equ 0x5C       ;temporary storage
0108
0109 ;      User Registers Bank 1
0110 ;W_TEMP equ 0xA0          ;Bank 1 temporary storage for W reg
0111
0112 ;      User defines
0113 #define tx    PORTB,7     ;define for RS232 TXD output pin
0114
0115 ;*****
0116
0117 org   0x0
0118 goto  init
0119

```

```

0004 28B9      0120    org    0x4
                0121    goto   ServiceInterruptions
                0122
                0123    org    0x10
0010           0124 init
0010 1283      0125    bcf   STATUS,RP0     ;select bank 0
0011 0185      0126    clrf   PORTA          ;clear PORTA and PORTB
0012 0186      0127    clrf   PORTB
0013 1786      0128    bsf   tx             ;set TXD output pin
0014 01A4      0129    clrf   flags           ;clear flags register
0015 3010      0130    movlw  0x10          ;load table offset register
0016 00A8      0131    movwf  offset
0017 018B      0132    clrf   INTCON         ;clear interrupt flags and disable interrupts
0018 3007      0133    movlw  0x07          ;turn off comparators, mode 111
0019 009F      0134    movwf  CMCON
001A 2140      0135    call   delay20       ;wait for comparators to settle
001B 089F      0136    movf   CMCON,F
001C 130C      0137    bcf   PIR1,CMIF
001D 1683      0138    bsf   STATUS,RP0     ;select bank 1
001E 3088      0139    movlw  0x88          ;WDT prescalar,internal TMR0 increment
001F 0081      0140    movwf  OPTION_REG
0020 0185      0141    clrf   TRISA          ;PORTA all outputs, discharges RC cks
0021 3060      0142    movlw  0x60          ;PORTA<7,4:0> outputs, PORTA<6:5> inputs
0022 0086      0143    movwf  TRISB
0023 300C      0144    movlw  0x0C          ;setup Voltage Reference
0024 009F      0145    movwf  VRCON
0025 1283      0146    bcf   STATUS,RP0     ;select bank 0
0026 3008      0147    movlw  0x08          ;enable RBIE interrupt
0027 008B      0148    movwf  INTCON
0028 213D      0149    call   vlong          ;delay before transmitting boot message
0029 213D      0150    call   vlong          ;to allow computer program to setup
002A 213D      0151    call   vlong
002B 2131      0152    call   BootMSG        ;transmit boot message
002C 178B      0153    bsf   INTCON,GIE    ;enable global interrupt bit
                0154
002D           0155 start
002D 1C24      0156    btfss  flags,BEGIN    ;wait for a start measurement key press
002E 282D      0157    goto   start
002F 1024      0158    bcf   flags,BEGIN    ;clear start measurement flag
                0159
0030 138B      0160    bcf   INTCON,GIE    ;transmit a start measurement message
0031 3053      0161    movlw  'S'           ;to the PC
0032 20AD      0162    call   Send
0033 178B      0163    bsf   INTCON,GIE
                0164
0034 01C2      0165    clrf   TimeHI         ;reset Time registers
0035 01C1      0166    clrf   TimeMID
0036 01C0      0167    clrf   TimeLO
0037 1E86      0168    btfss  PORTB,WHICH   ;detect if resistor or capacitor measure
0038 2862      0169    goto   Capacitor
                0170
0039           0171 Resistor
0039 1683      0172    bsf   STATUS,RP0     ;set V0 to input
003A 1406      0173    bsf   TRISB,V0
003B 1283      0174    bcf   STATUS,RP0
003C 20FB      0175    call   AnalogOn      ;turn analog on
003D 0181      0176    clrf   TMR0
003E 0000      0177    nop
003F 1586      0178    bsf   PORTB,V3       ;turn power on to unknown RC ckt
0040 19A4      0179 RwaitU   btfsc  flags,F_ERROR    ;detect if an error occurs
0041 288B      0180    goto   ErrorDetect
0042 1FA4      0181    btfss  flags,DONE      ;measurement completed flag
0043 2840      0182    goto   RwaitU
0044 13A4      0183    bcf   flags,DONE      ;clear measurement completed flag
0045 2111      0184    call   AnalogOff     ;turn analog off
                0185

```

```

0046 2126      0186    call   SwapTtoA      ;swap Time to accumulator a
0047 3007      0187    movlw  RkHI        ;swap known resistance value
0048 00D3      0188    movwf  ACCbHI      ;to accumulator b
0049 309D      0189    movlw  RkMID
004A 00D4      0190    movwf  ACCbMID
004B 3038      0191    movlw  RkLO
004C 00D5      0192    movwf  ACCbLO
004D 2230      0193    call   Mpy24       ;multiply accumulator a and b
0194
004E 1683      0195    bsf   STATUS,RP0     ;set V3 to input
004F 1586      0196    bsf   TRISB,V3
0050 1283      0197    bcf   STATUS,RP0
0051 20FB      0198    call   AnalogOn     ;turn analog on
0052 0181      0199    clrf  TMR0
0053 0000      0200    nop
0054 1406      0201    bsf   PORTB,V0      ;turn power on to known RC ckt
0055 19A4      0202    RwaitK btfsc flags,F_ERROR ;detect if an error occurs
0056 288B      0203    goto  ErrorDetect
0057 1FA4      0204    btfss flags,DONE     ;measurement completed flag

0058 2855      0205    goto  RwaitK
0059 13A4      0206    bcf   flags,DONE     ;clear measurement completed flag
005A 2111      0207    call   AnalogOff     ;turn analog off
0208
005B 2126      0209    call   SwapTtoA      ;swap Time to accumulator a
005C 224B      0210    call   Div24        ;divide multiply by known time
0211
005D 138B      0212    bcf   INTCON,GIE    ;disable all interrupts
005E 3052      0213    movlw  'R'          ;transmit, for R measurement
005F 20AD      0214    call   Send
0060 178B      0215    bsf   INTCON,GIE    ;enable global interrupt bit
0061 282D      0216    goto  start         ;restart
0217
0062 1683      0218    Capacitor
0062 1683      0219    bsf   STATUS,RP0     ;set V2 to input
0063 1506      0220    bsf   TRISB,V2
0064 1283      0221    bcf   STATUS,RP0
0065 20FB      0222    call   AnalogOn     ;turn analog on
0066 0181      0223    clrf  TMR0
0067 0000      0224    nop
0068 1486      0225    bsf   PORTB,V1      ;turn power on to unknown RC ckt
0069 19A4      0226    CwaitU btfsc flags,F_ERROR ;detect if an error occurs
006A 288B      0227    goto  ErrorDetect
006B 1FA4      0228    btfss flags,DONE     ;measurement completed flag
006C 2869      0229    goto  CwaitU
006D 13A4      0230    bcf   flags,DONE     ;clear measurement completed flag
006E 2111      0231    call   AnalogOff     ;turn analog off
0232
006F 2126      0233    call   SwapTtoA      ;swap Time to accumulator a
0070 3007      0234    movlw  CkHI        ;swap known resistance value
0071 00D3      0235    movwf  ACCbHI      ;to accumulator b
0072 30C8      0236    movlw  CkMID
0073 00D4      0237    movwf  ACCbMID
0074 3030      0238    movlw  CkLO
0075 00D5      0239    movwf  ACCbLO
0076 2230      0240    call   Mpy24       ;multiply accumulator a and b
0241
0077 1683      0242    bsf   STATUS,RP0     ;set V3 to input
0078 1606      0243    bsf   TRISB,V4
0079 1283      0244    bcf   STATUS,RP0
007A 20FB      0245    call   AnalogOn     ;turn analog on
007B 0181      0246    clrf  TMR0
007C 0000      0247    nop
007D 1486      0248    bsf   PORTB,V1      ;turn power on to known RC ckt
007E 19A4      0249    CwaitK btfsc flags,F_ERROR ;detect if an error occurs
007F 288B      0250    goto  ErrorDetect

```

```

0080 1FA4      0251    btfss   flags,DONE      ;measurement completed flag
0081 287E      0252    goto    CwaitK
0082 13A4      0253    bcf     flags,DONE      ;clear measurement completed flag
0083 2111      0254    call    AnalogOff      ;turn analog off
0255
0084 2126      0256    call    SwapTtoA      ;swap Time to accumulator a
0085 224B      0257    call    Div24        ;divide multiply by known time
0258
0086 138B      0259    bcf     INTCON,GIE    ;disable all interrupts
0087 3043      0260    movlw   'C'          ;transmit, for C measurement
0088 20AD      0261    call    Send
0089 178B      0262    bsf    INTCON,GIE    ;enable global interrupt bit
008A 282D      0263    goto   start        ;restart
0264
008B           0265    ErrorDetect
008B 1283      0266    bcf    STATUS,RP0      ;disable TMR0
008C 128B      0267    bcf    INTCON,T0IE
008D 110B      0268    bcf    INTCON,T0IF
008E 2111      0269    call   AnalogOff      ;turn analog off
008F 11A4      0270    bcf    flags,F_ERROR  ;clear error flag
0271
0090 138B      0272    bcf    INTCON,GIE    ;disable all interrupts
0091 3045      0273    movlw   'E'          ;transmit, for C measurement
0092 20AD      0274    call   Send
0093 178B      0275    bsf    INTCON,GIE    ;enable global interrupt bit
0094 282D      0276    goto   start        ;restart
0277
0278 ;*****
0279 ;-----
0280 ;*- RS232 Transmit Routine
0281 ;*- Borrowed from AN593, "Serial Port Routines Without Using the RTC"
0282 ;*- Author: Stan D'Souza
0283 ;*- This is the routine that interfaces directly to the hardware
0284 ;-----*
0285 ;*****
0095           0286    Transmit
0095 1283      0287    bcf    STATUS,RP0
0096 00A6      0288    movwf  txreg
0097 1386      0289    bcf    tx          ;send start bit
0098 3020      0290    movlw   baudconst
0099 00A7      0291    movwf  delay
009A 3009      0292    movlw   0x9
009B 00A5      0293    movwf  count
009C
0294 txbaudwait
009C 0BA7      0295    decfsz delay
009D 289C      0296    goto   txbaudwait
009E 3020      0297    movlw   baudconst
009F 00A7      0298    movwf  delay
00A0 0BA5      0299    decfsz count
00A1 28A6      0300    goto   SendNextBit
00A2 3009      0301    movlw   0x9
00A3 00A5      0302    movwf  count
00A4 1786      0303    bsf    tx          ;send stop bit
00A5 0008      0304    return
00A6
0305 SendNextBit
00A6 0CA6      0306    rrf    txreg
00A7 1C03      0307    btfss STATUS,C
00A8 28AB      0308    goto   Setlo
00A9 1786      0309    bsf    tx
00AA 289C      0310    goto   txbaudwait
00AB 1386      0311    Setlo  bcf    tx
00AC 289C      0312    goto   txbaudwait
0313 ;
0314
0315 ;*****
0316 ;-----

```

```

0317 ;*- Generic Transmit Routine
0318 ;*- Sends what is currently in the W register and accumulator ACCc
0319 ;*-----
0320 ;*****-----*****-----*****-----*****-----*****-----*****-----*
00AD 00AD Send
00AD 2095 0321 call Transmit
00AE 2146 0322 call delay1 ;delay between bytes
00AF 0856 0323 movf ACCcHI,W ;transmit high resistance byte
00B0 2095 0324 call Transmit
00B1 2146 0325 call delay1 ;delay between bytes
00B2 0857 0326 movf ACCcMID,W ;transmit mid resistance byte
00B3 2095 0327 call Transmit
00B4 2146 0328 call delay1 ;delay between bytes
00B5 0858 0329 movf ACCcLO,W ;transmit low resistance byte
00B6 2095 0330 call Transmit
00B7 2146 0331 call delay1 ;delay between bytes
00B8 0008 0332 return
0333 ;-----
0334 ;-----
0335 ;-----
0336 ;*****-----*****-----*****-----*****-----*****-----*****-----*
0337 ;*-----
0338 ;*- Interrupt Service Routines
0339 ;*-----
0340 ;*****-----*****-----*****-----*****-----*****-----*****-----*
00B9 00A0 ServiceInterrupts
00B9 00A0 0341 movwf W_TEMP ;Pseudo push instructions
00BA 0E03 0342 swapf STATUS,W
00BB 1283 0343 bcf STATUS,RPO
00BC 00A1 0344 movwf STATUS_TEMP
0345
00BD 0801 0346
00BE 00A3 0347 movf TMR0,W
00BF 190B 0348 movwf Ttemp
00C0 20E5 0349 btfsc INTCON,T0IF ;Service Timer 0 overflow
00C1 1B0C 0350 call ServiceTimer
00C2 20EC 0351 btfsc PIR1,CMIF ;Stops Timer0, Records Value
00C3 180B 0352 call ServiceComparator
00C4 20CB 0353 btfsc INTCON,RBIF ;Service pushbutton switch
0354 call ServiceKeystroke ;Starts a measurement
0355
00C5 1283 0356 bcf STATUS,RPO
00C6 0E21 0357 swapf STATUS_TEMP,W ;Pseudo pop instructions
00C7 0083 0358 movwf STATUS
00C8 0EA0 0359 swapf W_TEMP,F
00C9 0E20 0360 swapf W_TEMP,W
0361
00CA 0009 0362 retfie
0363 ;-----
0364 ;-----
0365 ;*****-----*****-----*****-----*****-----*****-----*****-----*
0366 ;*-----
0367 ;*- Borrowed from AN552, "Implementing Wake-up on Key Stroke"
0368 ;*- Author: Stan D'Souza
0369 ;*-----
0370 ;*****-----*****-----*****-----*****-----*****-----*****-----*
00CB 00CB ServiceKeystroke
00CB 118B 0371 bcf INTCON,RBIE ;disable interrupt
00CC 0906 0372 comf PORTB,W ;read PORTB
00CD 100B 0373 bcf INTCON,RBIF ;clear interrupt flag
00CE 3940 0374 andlw B'01000000'
0375
00CF 1903 0376 btfsc STATUS,Z
00D0 28D6 0377 goto NotSwitch
00D1 2143 0378 call delay16 ;de-bounce switch for 16msec
00D2 0906 0379 comf PORTB,W ;read PORTB again
00D3 20D9 0380 call KeyRelease ;check for key release
00D4 1424 0381 bsf flags,BEGIN
00D5 0008 0382 return

```

```

0383
00D6          0384 NotSwitch           ;detected other PORTB pin change
00D6 100B      0385     bcf    INTCON,RBIF   ;reset RBI interrupt
00D7 158B      0386     bsf    INTCON,RBIE
00D8 0008      0387     return
0388
00D9          0389 KeyRelease
00D9 2143      0390     call   delay16    ;debounce switch
00DA 0906      0391     comf   PORTB,W    ;read PORTB
00DB 100B      0392     bcf    INTCON,RBIF   ;clear flag
00DC 158B      0393     bsf    INTCON,RBIE  ;enable interrupt
00DD 3940      0394     andlw B'01000000'
00DE 1903      0395     btfsc  STATUS,Z   ;key still pressed?
00DF 0008      0396     return
00E0 0063      0397     sleep
00E1 118B      0398     bcf    INTCON,RBIE  ;disable interrupts
00E2 0906      0399     comf   PORTB,W    ;read PORTB
00E3 100B      0400     bcf    INTCON,RBIF   ;clear flag
00E4 28D9      0401     goto   KeyRelease ;try again
0402 ;
0403
0404 ;*****
0405 ;*-----*
0406 ;*- ISR to service a Timer0 overflow
0407 ;*-----*
0408 ;*****
00E5          0409 ServiceTimer
00E5 0AC1      0410     incf   TimeMID,F  ;increment middle Time byte
00E6 1903      0411     btfsc  STATUS,Z   ;if middle overflows,
00E7 0AC2      0412     incf   TimeHI,F   ;increment high Time byte
00E8 1AC2      0413     btfsc  TimeHI,EMPTY ;check if component is connected
00E9 15A4      0414     bsf    flags,F_ERROR ;set error flag
00EA 110B      0415     bcf    INTCON,T0IF   ;clear TMR0 interrupt flag
00EB 0008      0416     return
0417 ;
0418
0419 ;*****
0420 ;*-----*
0421 ;*- ISR to service a Comparator interrupt
0422 ;*-----*
0423 ;*****
00EC          0424 ServiceComparator
00EC 1283      0425     bcf    STATUS,RP0   ;select bank 0
00ED 1E86      0426     btfss  PORTB,WHICH ;detect which measurement, R or C?
00EE 28F2      0427     goto   capcomp
00EF 1F1F      0428     btfss  CMCON,C1OUT ;detect if R ckt has interrupted
00F0 28F4      0429     goto   scstop
00F1 28F9      0430     goto   scend
00F2          0431 capcomp
00F2 1B9F      0432     btfsc  CMCON,C2OUT ;detect if C ckt has interrupted
00F3 28F9      0433     goto   scend
00F4          0434 scstop
00F4 128B      0435     bcf    INTCON,T0IE   ;disable TMR0 interrupts
00F5 110B      0436     bcf    INTCON,T0IF
00F6 0823      0437     movf   Ttemp,W
00F7 00C0      0438     movwf  TimeLO
00F8 17A4      0439     bsf    flags,DONE   ;set DONE flag
00F9          0440 scend
00F9 130C      0441     bcf    PIR1,CMIF   ;clear comparator interrupt flag
00FA 0008      0442     return
0443 ;
0444
0445 ;*****
0446 ;*-----*
0447 ;*- Turn Comparators and Vref On
0448 ;*-----*

```

```

0449 ;*****
00FB 0450 AnalogOn
00FB 1283 0451 bcf STATUS,RP0 ;select bank 0
00FC 3002 0452 movlw 0x02 ;turn comparators on, mode 010
00FD 009F 0453 movwf CMCON ;4 inputs multiplexed to 2 comparators
00FE 1683 0454 bsf STATUS,RP0 ;select bank 1
00FF 300F 0455 movlw 0x0F ;make PORTA<3:0> all inputs
0100 0085 0456 movwf TRISA
0101 179F 0457 bsf VRCON,VREN
0102 1283 0458 bcf STATUS,RP0 ;select bank 0
0103 2140 0459 call delay20 ;20msec delay
0104 089F 0460 movf CMCON,F ;clear comparator mismatch condition
0105 130C 0461 bcf PIR1,CMIF ;clear comparator interrupt flag
0106 1683 0462 bsf STATUS,RP0
0107 170C 0463 bsf PIE1,CMIE ;enable comparator interrupts
0108 1283 0464 bcf STATUS,RP0
0109 170B 0465 bsf INTCON,PEIE ;enable peripheral interrupts
010A 11A4 0466 bcf flags,F_ERROR
010B 0181 0467 clrf TMR0 ;clear TMR0 counter
010C 0000 0468 nop
010D 0000 0469 nop
010E 110B 0470 bcf INTCON,TOIF ;clear TMR0 interrupt flag
010F 168B 0471 bsf INTCON,TOIE ;enable TMR0 interrupts
0110 0008 0472 return
0473 ;-----
0474
0475 ;*****
0476 ;*-----*
0477 ;*- Turn Comparators and Vref Off
0478 ;*-----*
0479 ;*****
0111 0480 AnalogOff
0111 1283 0481 bcf STATUS,RP0
0112 130B 0482 bcf INTCON,PEIE
0113 3080 0483 movlw 0x80 ;reset PORTB value
0114 0086 0484 movwf PORTB
0115 1683 0485 bsf STATUS,RP0 ;select bank 1
0116 130C 0486 bcf PIE1,CMIE ;disable comparator interrupts
0117 0185 0487 clrf TRISA ;set PORTA pins to outputs, discharge RC ckt
0118 3060 0488 movlw 0x60 ;set PORTB 7,4-0 as outputs, 6,5 as inputs
0119 0086 0489 movwf TRISB
011A 139F 0490 bcf VRCON,VREN ;disable Vref
011B 1283 0491 bcf STATUS,RP0 ;select bank 0
011C 3007 0492 movlw 0x07
011D 009F 0493 movwf CMCON ;disable comparators
011E 2140 0494 call delay20 ;20msec delay
011F 089F 0495 movf CMCON,F ;clear comparator mismatch condition
0120 130C 0496 bcf PIR1,CMIF ;clear comparator interrupt flag
0121 110B 0497 bcf INTCON,TOIF ;clear Timer0 interrupt flag
0122 213D 0498 call vlong ;long delay to allow capacitors to discharge
0123 213D 0499 call vlong
0124 213D 0500 call vlong
0125 0008 0501 return
0502 ;-----
0503
0504 ;*****
0505 ;*-----*
0506 ;*- Swap Time to Accumulator a
0507 ;*-----*
0508 ;*****
0126 0509 SwapTtoa
0126 1283 0510 bcf STATUS,RP0
0127 0842 0511 movf TimeHI,W
0128 00D0 0512 movwf ACCaHI
0129 0841 0513 movf TimeMID,W
012A 00D1 0514 movwf ACCaMID

```

```

012B 0840      0515      movf    TimeLO,W
012C 00D2      0516      movwf   ACCaLO
012D 01C2      0517      clrf    TimeHI
012E 01C1      0518      clrf    TimeMID
012F 01C0      0519      clrf    TimeLO
0130 0008      0520      return
0521 ;
0522
0523 ;*****
0524 ;-----
0525 ;*-      Transmit the Boot Message
0526 ;-----
0527 ;*****
0131          0528 BootMSG
0131 1283      0529      bcf     STATUS,RP0      ;select bank 0
0132 3002      0530 msg     movlw   HIGH Table    ;init the PCH for a table call
0133 008A      0531      movwf   PCLATH
0134 0828      0532      movf    offset,W      ;move table offset into W
0135 2200      0533      call    Table        ;get table value
0136 2095      0534      call    Transmit      ;transmit table value
0137 2146      0535      call    delay1       ;delay between bytes
0138 0BA8      0536      decfsz offset,F      ;check for end of table
0139 2932      0537      goto   msg
013A 3010      0538      movlw   0x10          ;reset table offset
013B 00A8      0539      movwf   offset
013C 0008      0540      return
0541 ;
0542
0543 ;*****
0544 ;-----
0545 ;*-      Delay Routines
0546 ;-----
0547 ;*****
013D 30FF      0548 vlong   movlw   0xff          ;very long delay, approx 200msec
013E 00A9      0549      movwf   msb
013F 2948      0550      goto   d1
0140          0551 delay20           ;20 msec delay
0140 301A      0552      movlw   .26
0141 00A9      0553      movwf   msb
0142 2948      0554      goto   d1
0143          0555 delay16           ;16 msec delay
0143 3015      0556      movlw   .21
0144 00A9      0557      movwf   msb
0145 2948      0558      goto   d1
0146          0559 delay1           ;approx 750nsec delay
0146 3001      0560      movlw   .1
0147 00A9      0561      movwf   msb
0148 30FF      0562 d1      movlw   0xff
0149 00AA      0563      movwf   lsb
014A 0BAA      0564 d2      decfsz lsb,F
014B 294A      0565      goto   d2
014C 0BA9      0566      decfsz msb,F
014D 2948      0567      goto   d1
014E 0008      0568      return
0569 ;
0570
0571
0572      org     0x200
0573
0574
0575 ;*****
0576 ;-----
0577 ;*-      Table for Boot Message
0578 ;-----
0579 ;*****
0200          0580 Table           ;boot message "PICMETER Booted!"

```

```
0200 0782      0581      addwf   PCL          ;add W to PCL
0201 3400      0582      retlw   0
0202 3421      0583      retlw   '!'
0203 3464      0584      retlw   'd'
0204 3465      0585      retlw   'e'
0205 3474      0586      retlw   't'
0206 346F      0587      retlw   'o'
0207 346F      0588      retlw   'o'
0208 3442      0589      retlw   'B'
0209 3420      0590      retlw   ' '
020A 3472      0591      retlw   'r'
020B 3465      0592      retlw   'e'
020C 3474      0593      retlw   't'
020D 3465      0594      retlw   'e'
020E 346D      0595      retlw   'm'
020F 3443      0596      retlw   'C'
0210 3449      0597      retlw   'I'
0211 3450      0598      retlw   'P'
0599 ;
0600
0601 ;*****
0602 ;-----
0603 ;*-    24-bit Addition
0604 ;*-
0605 ;*-    Uses ACCa and ACCb
0606 ;*-
0607 ;*-    ACCa + ACCb -> ACCb
0608 ;-----
0609 ;*****
0212
0212 0852      0611      movf    ACCaLO,W
0213 07D5      0612      addwf   ACCbLO          ;add low bytes
0214 1803      0613      btfsc  STATUS,C        ;add in carry if necessary
0215 2A1D      0614      goto   A2
0216 0851      0615 A1     movf    ACCaMID,W
0217 07D4      0616      addwf   ACCbMID         ;add mid bytes
0218 1803      0617      btfsc  STATUS,C        ;add in carry if necessary
0219 0AD3      0618      incf    ACCbHI
021A 0850      0619      movf    ACCaHI,W
021B 07D3      0620      addwf   ACCbHI          ;add high bytes
021C 3400      0621      retlw   0
021D 0AD4      0622 A2     incf    ACCbMID
021E 1903      0623      btfsc  STATUS,Z
021F 0AD3      0624      incf    ACCbHI
0220 2A16      0625      goto   A1
0626 ;
0627
0628 ;*****
0629 ;-----
0630 ;*-    Subtraction ( 24 - 24 -> 24 )
0631 ;*-
0632 ;*-    Uses ACCa, ACCb, ACCd
0633 ;*-
0634 ;*-    ACCa -> ACCd,
0635 ;*-    2's complement ACCa,
0636 ;*-    call Add24 ( ACCa + ACCb -> ACCb ),
0637 ;*-    ACCd -> ACCa
0638 ;-----
0639 ;*****
0221
0221 0850      0640 Sub24
0641      movf    ACCaHI,W          ;Transfer ACCa to ACCd
0222 00D9      0642      movwf   ACCdHI
0223 0851      0643      movf    ACCaMID,W
0224 00DA      0644      movwf   ACCdMID
0225 0852      0645      movf    ACCaLO,W
0226 00DB      0646      movwf   ACCdLO
```

```

0227 2275      0647    call     compA          ;2's complement ACCa
0228 2212      0648    call     Add24         ;Add ACCa to ACCb
0229 0859      0649    movf    ACCdHI,W       ;Transfer ACCd to ACCa
022A 00D0      0650    movwf   ACCaHI
022B 085A      0651    movf    ACCdMID,W
022C 00D1      0652    movwf   ACCaMID
022D 085B      0653    movf    ACCdLO,W
022E 00D2      0654    movwf   ACCaLO
022F 3400      0655    retlw   0
0656 ;
0657
0658 ;*****
0659 ;-----
0660 ;*-      Multiply ( 24 X 24 -> 56 )
0661 ;*-      Uses ACCa, ACCb, ACCc, ACCd
0663 ;*-
0664 ;*-      ACCa * ACCb -> ACCb,ACCc 56-bit output
0665 ;*-      with ACCb (ACCbHI,ACCbMID,ACCbLO) with 24 msb's and
0666 ;*-      ACCc (ACCcHI,ACCcMID,ACCcLO) with 24 lsb's
0667 ;*-----
0668 ;*****
0230      0669 Mpy24
0230 223F      0670    call     Msetup
0231 0CD9      0671    mloop   rrf    ACCdHI          ;rotate d right
0232 0CDA      0672    rrf    ACCdMID
0233 0CDB      0673    rrf    ACCdLO
0234 1803      0674    btfsc  STATUS,C        ;need to add?
0235 2212      0675    call     Add24
0236 0CD3      0676    rrf    ACCbHI
0237 0CD4      0677    rrf    ACCbMID
0238 0CD5      0678    rrf    ACCbLO
0239 0CD6      0679    rrf    ACCcHI
023A 0CD7      0680    rrf    ACCcMID
023B 0CD8      0681    rrf    ACCcLO
023C 0BDC      0682    decfsz temp           ;loop until all bits checked
023D 2A31      0683    goto    mloop
023E 3400      0684    retlw   0
0685
023F      0686 Msetup
023F 3018      0687    movlw   0x18          ;for 24 bit shifts
0240 00DC      0688    movwf   temp
0241 0853      0689    movf    ACCbHI,W       ;move ACCb to ACCd
0242 00D9      0690    movwf   ACCdHI
0243 0854      0691    movf    ACCbMID,W
0244 00DA      0692    movwf   ACCdMID
0245 0855      0693    movf    ACCbLO,W
0246 00DB      0694    movwf   ACCdLO
0247 01D3      0695    clrf    ACCbHI
0248 01D4      0696    clrf    ACCbMID
0249 01D5      0697    clrf    ACCbLO
024A 3400      0698    retlw   0
0699 ;
0700
0701 ;*****
0702 ;-----
0703 ;*-      Division ( 56 / 24 -> 24 )
0704 ;*-      Uses ACCa, ACCb, ACCc, ACCd
0706 ;*-
0707 ;*-      56-bit dividend in ACCb,ACCc ( ACCb has msb's and ACCc has lsb's)
0708 ;*-      24-bit divisor in ACCa
0709 ;*-      quotient is stored in ACCc
0710 ;*-      remainder is stored in ACCb
0711 ;*-----
0712 ;*****

```

```

024B          0713 Div24
024B 2272    0714     call   Dsetup
0715
024C 1003    0716 dloop  bcf    STATUS,C
024D 0DD8    0717     rlf   ACCcLO      ;Rotate dividend left 1 bit position
024E 0DD7    0718     rlf   ACCcMID
024F 0DD6    0719     rlf   ACCcHI
0250 0DD5    0720     rlf   ACCbLO
0251 0DD4    0721     rlf   ACCbMID
0252 0DD3    0722     rlf   ACCbHI
0723
0253 1803    0724     btfsc  STATUS,C      ;invert carry and exclusive or with the
0254 2A58    0725     goto   clear        ;msb of the divisor then move this bit
0255 1FD0    0726     btfss  ACCaHI,msb_bit ;into the lsb of the dividend
0256 0AD8    0727     incf   ACCcLO
0257 2A5A    0728     goto   cont
0258 1BD0    0729     clear  btfsc  ACCaHI,msb_bit
0259 0AD8    0730     incf   ACCcLO
0731
025A 1858    0732     cont   btfsc  ACCcLO,lsb_bit ;check the lsb of the dividend
025B 2A5E    0733     goto   minus
025C 2212    0734     call   Add24      ;if = 0, then add divisor to upper 24 bits
025D 2A5F    0735     goto   check       ;of dividend
025E 2221    0736     minus  call   Sub24      ;if = 1, then subtract divisor from upper
0737           24 bits of dividend
0738
025F 0BDC    0739     check  decfsz temp,f      ;do 24 times
0260 2A4C    0740     goto   dloop
0741
0261 1003    0742     bcf   STATUS,C
0262 0DD8    0743     rlf   ACCcLO      ;shift lower 24 bits of dividend 1 bit
0263 0DD7    0744     rlf   ACCcMID
0264 0DD6    0745     rlf   ACCcHI
0265 1BD3    0746     btfsc ACCbHI,msb_bit ;exclusive or the inverse of the msb of the
0266 2A6A    0747     goto   w1        ;dividend with the msb of the divisor
0267 1FD0    0748     btfss ACCaHI,msb_bit ;store in the lsb of the dividend
0268 0AD8    0749     incf   ACCcLO
0269 2A6C    0750     goto   wzd
026A 1BD0    0751     w1     btfsc ACCaHI,msb_bit
026B 0AD8    0752     incf   ACCcLO
026C 1FD3    0753     wzd   btfss ACCbHI,msb_bit ;if the msb of the remainder is set and
026D 2A71    0754     goto   wend
026E 1BD0    0755     btfsc ACCaHI,msb_bit ;the msb of the divisor is not
026F 2A71    0756     goto   wend
0270 2212    0757     call   Add24      ;add the divisor to the remainder to correct
0758           24 bits of dividend
0759           ;for zero partial remainder
0271 3400    0760     wend  retlw 0      ;quotient in 24 lsb's of dividend
0761           0762           ;remainder in 24 msb's of dividend
0763 Dsetup
0272 3018    0764     movlw  0x18      ;loop 24 times
0273 00DC    0765     movwf  temp
0766
0274 3400    0767     retlw  0
0768 ;
0769
0770 ;*****-*-----*
0771 ;*------*
0772 ;*-      2's Complement
0773 ;*------*
0774 ;*-      Uses ACCa
0775 ;*------*
0776 ;*-      Performs 2's complement conversion on ACCa
0777 ;*------*
0778 ;*****-*-----*

```

---

---

```
0275          0779 compA
0275 09D2      0780      comf   ACCaLO      ;invert all bits in accumulator a
0276 09D1      0781      comf   ACCaMID
0277 09D0      0782      comf   ACCaHI
0278 0AD2      0783      incf   ACCaLO      ;add one to accumulator a
0279 1903      0784      btfsc  STATUS,Z
027A 0AD1      0785      incf   ACCaMID
027B 1903      0786      btfsc  STATUS,Z
027C 0AD0      0787      incf   ACCaHI
027D 3400      0788      retlw  0
0789 ;_____
0790
0791      END
0792
```

---

```
0000 : ----- XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
0040 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
0080 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
00C0 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
0100 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
0140 : XXXXXXXXXXXXXXXXX- ----- -----
0200 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX
0240 : XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXXX XXXXXXXXXXXXXXXXX--
```

All other memory blocks unused.

```
Errors   :    0
Warnings :    0
Messages :    0
```

Please check the Microchip BBS for the latest version of the source code. For BBS access information, see Section 6, Microchip Bulletin Board Service, page 6-3.

## APPENDIX B: VISUAL BASIC PROGRAM

### PICMTR.FRM

```
Sub Form_Load ()  
    'Initialize the program  
    Image1.Height = 600  
    Image1.Width = 2700  
    Frame1.Caption = "PICMETER Power Off"  
    Label1.Caption = ""  
    Label2.Caption = ""  
  
    'Initialize Comm Port 1  
    Comm1.RThreshold = 1  
    Comm1.Handshaking = 0  
    Comm1.Settings = "9600,n,8,1"  
    Comm1.CommPort = 2  
    Comm1.PortOpen = True  
  
    'Initialize the global variable First%  
    First% = 0  
End Sub  
  
Sub Form_Unload (Cancel As Integer)  
    'Unload PICMETER  
    Comm1.RTSEnable = False  
    Comm1.DTREnable = False  
    Comm1.PortOpen = False  
    Unload PICMETER  
End Sub  
  
Sub Comm1_OnComm ()  
    Dim Value As Double  
    Dim High As Double  
    Dim Medium As Double  
    Dim Low As Double  
  
    'Received a character  
    If Comm1.CommEvent = 2 Then  
        If First% = 0 Then  
            If Comm1.InBufferCount = 16 Then  
                Label1.FontSize = 10  
                InString$ = Comm1.Input  
                If InString$ = "PICMETER Booted!" Then  
                    Frame1.Caption = "PICMETER Booted!"  
                End If  
                First% = 1  
                Comm1.InputLen = 4  
            End If  
        Else  
            If Comm1.InBufferCount >= 4 Then  
                InString$ = Comm1.Input  
                If Left$(InString$, 1) = "R" Then  
                    Frame1.Caption = "Resistance"  
                    Label2.FontName = "Symbol"  
                    Label2.Caption = "KW"  
                    Label1.FontSize = 24  
                ElseIf Left$(InString$, 1) = "C" Then  
                    Frame1.Caption = "Capacitance"  
                    Label2.FontName = "MS Sans Serif"  
                    Label2.Caption = "nF"  
                    Label1.FontSize = 24  
                ElseIf Left$(InString$, 1) = "E" Then  
                    Frame1.Caption = "Error Detected"  
                    Label2.Caption = ""  
                ElseIf Left$(InString$, 1) = "S" Then  
                    Frame1.Caption = "Measuring Component"  
            End If  
        End If  
    End If  
End Sub
```

```

        Label2.Caption = ""
    Else
        Frame1.Caption = "Error Detected"
        Label2.Caption = ""
    End If

    If Frame1.Caption = "Error Detected" Then
        Label1.Caption = ""
    ElseIf Frame1.Caption = "Measuring Component" Then
        Label1.Caption = ""
    Else
        High = 65536# * Asc(Mid$(InString$, 2, 1))
        Medium = 256# * Asc(Mid$(InString$, 3, 1))
        Low = Asc(Mid$(InString$, 4, 1))
        Label1.Caption = Format$((High + Medium + Low) / 1000, "###0.0")
    End If
    End If
End If
End Sub

Sub Check3D1_Click (Value As Integer)
'Control Power to the PICMETER
If Check3D1.Value = False Then
    Comm1.InputLen = 0
    Label1.Caption = ""
    Label2.Caption = ""
    Comm1.RTSEnable = False
    Comm1.DTREnable = False
    Frame1.Caption = "PICMETER Power Off"
    InString$ = Comm1.Input
Else
    Frame1.Caption = ""
    First% = 0
    Comm1.InputLen = 0
    InString$ = Comm1.Input
    Comm1.RTSEnable = True
    Comm1.DTREnable = True
End If
End Sub

Sub menExitTop_Click ()
'Unload PICMETER
Unload PICMETER
End Sub

Sub Option1_Click ()
'Open COM1 for communications
If Option1.Value = True Then
    If Comm1.CommPort = 2 Then
        Comm1.PortOpen = False
        Comm1.CommPort = 1
        Comm1.PortOpen = True
    End If
End If
End Sub

Sub Option2_Click ()
'Open COM2 for communications
If Option2.Value = True Then
    If Comm1.CommPort = 1 Then
        Comm1.PortOpen = False
        Comm1.CommPort = 2
        Comm1.PortOpen = True
    End If
End If
End Sub

```

**PICMETER.BAS**

```

Global I%
Global First%

```

# AN611

## APPENDIX C: PICMETER PCB LAYOUT

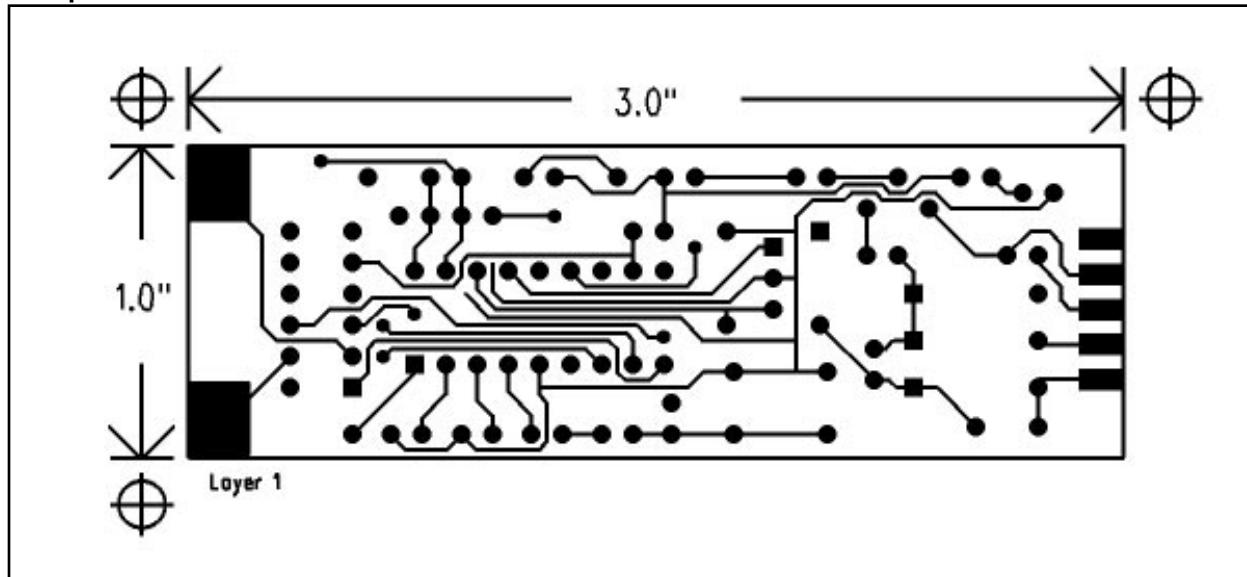
Boards Manufactured by:

Southwest Circuits

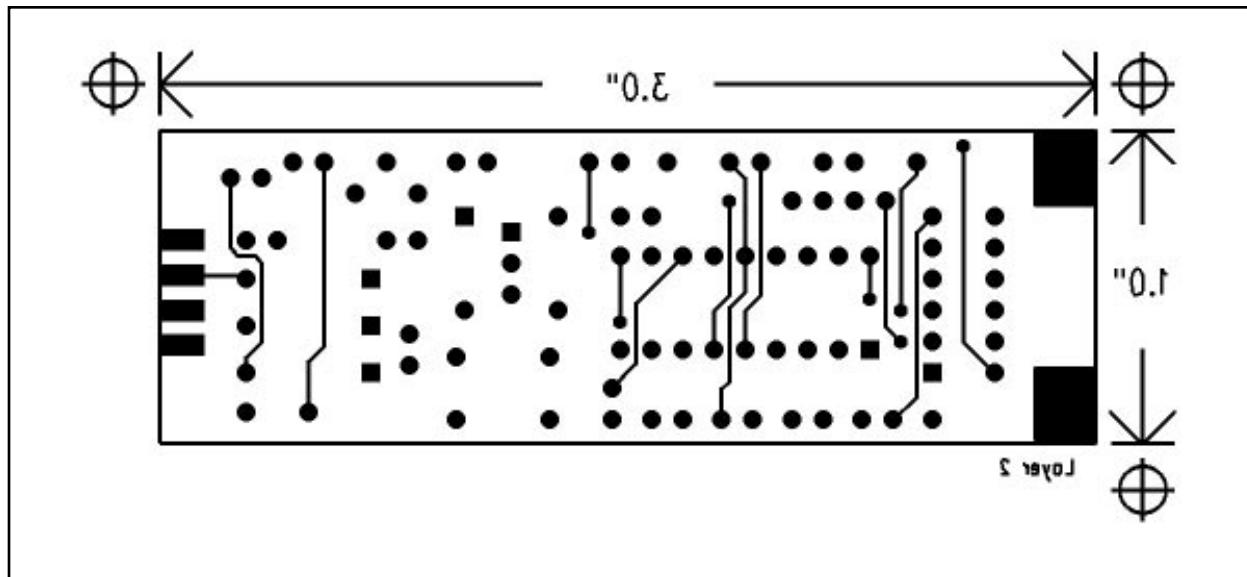
Contact: Perry Groves  
3760 E. 43rd Place  
Tucson, AZ 85713  
1-520-745-8515

The following artwork is not printed to scale:

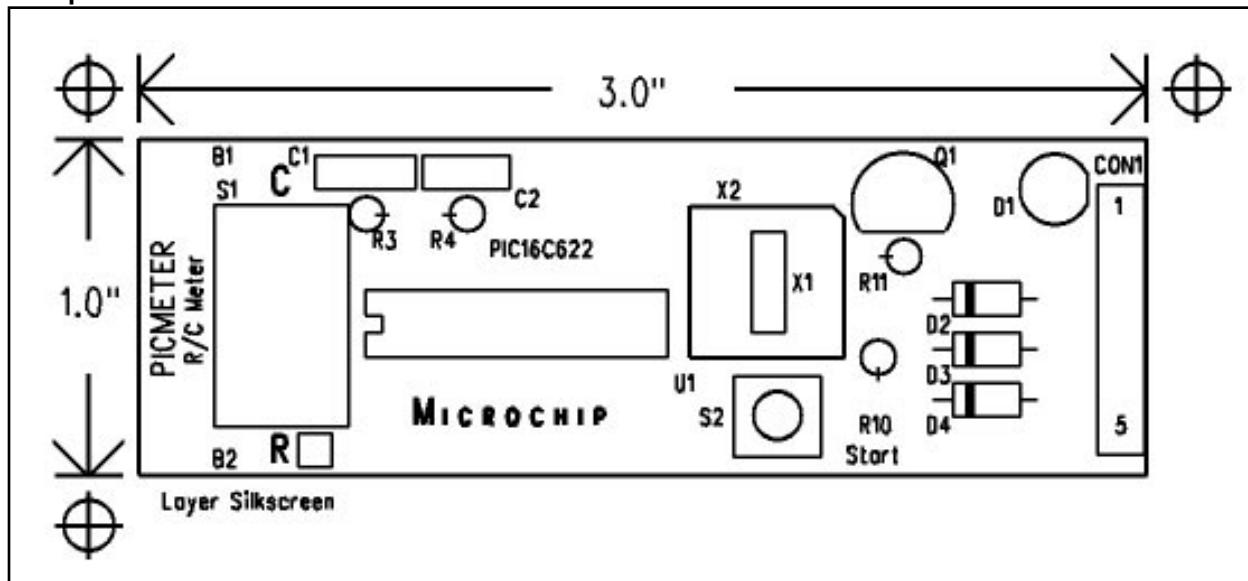
**Component Side**



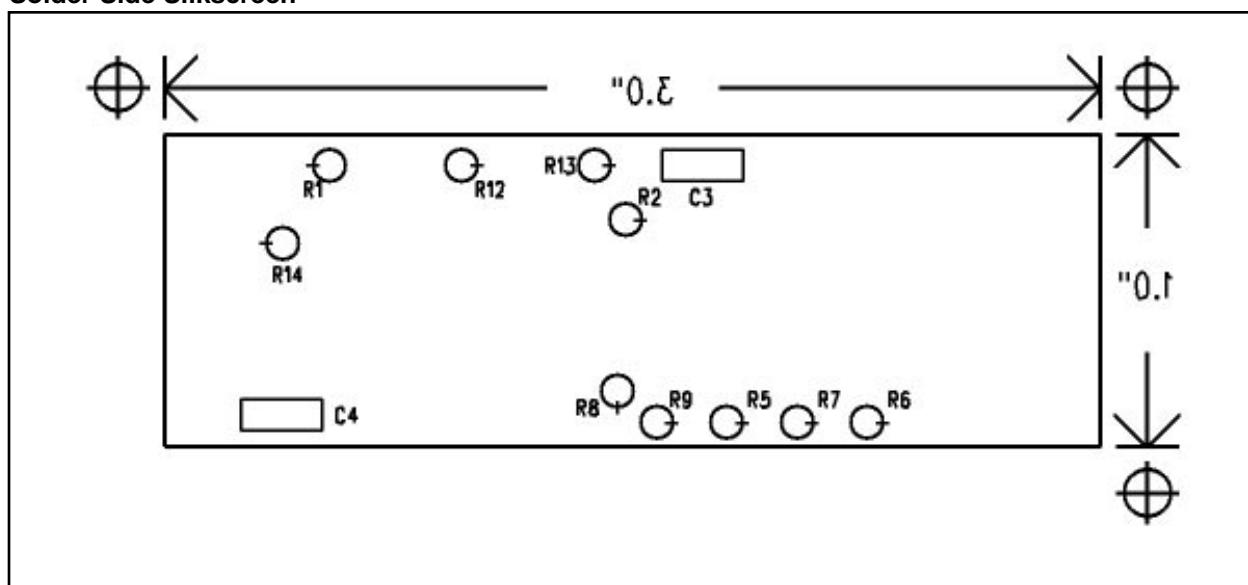
**Solder Side**



## Component Side Silkscreen



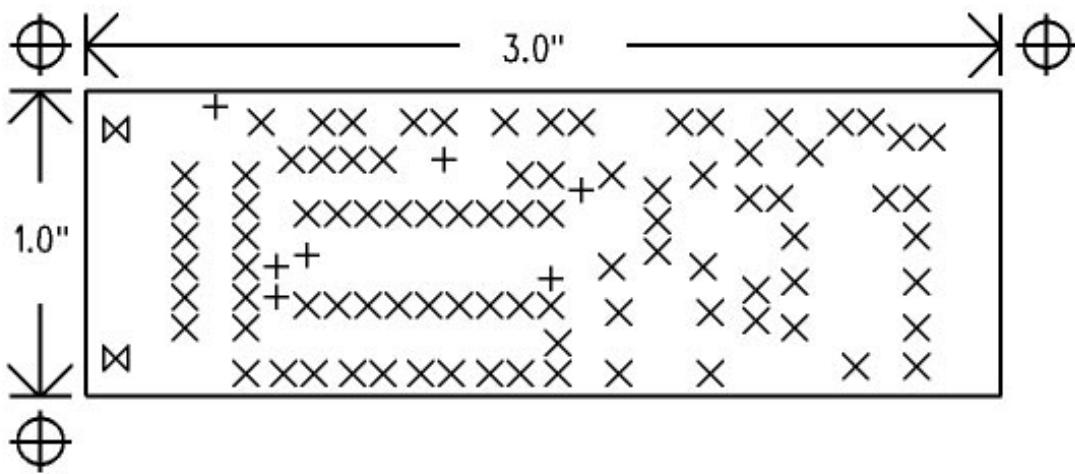
## Solder Side Silkscreen



# AN611

---

## Manufacturing Drawing



SIZE	QTY	SYM
18	7	+
37	89	X
95	2	⊗

**NOTES:**

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